# **Power Systems Analysis** Chapter 2 Transmission line models

## Outline

- 1. Line characteristics
- 2. Line models

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#### 1. Line characteristics

- Resistance *r* and conductance *g*
- Series inductance *l*
- Shunt capacitance *c*
- Balanced  $3\phi$  lines

#### 2. Line models

# **3***\phi* line

Alternating currents in conductors line interact electromagnetically Interactions couple voltages & currents across phases In balanced operation, phases behave as if they are decoupled In each phase, line is characterized by

- series impedance / meter
- $\Omega^{-1}/m$ • shunt admittance / meter to neutral  $y := g + i\omega c$

- $\Omega/m$  $z := r + i\omega l$

## Assumption

Currents and charges sum to zero across all *n* conductors:

- $i_1(t) + \cdots + i_p$
- $q_1(t) + \dots + q_n(t)$

$$f_n(t) = 0$$
 for all  $t$ 

$$q_n(t) = 0$$
 for all  $t$ 

- 1. Series inductance *l* 
  - total flux linkages  $\lambda_k$  of conductor k depends on all currents  $i_{k'}$

$$h_k = \left(\frac{\mu_0}{2\pi} \ln \frac{1}{r'_k}\right) i_k +$$

self inductance henrys/m

 $\sum_{\substack{k' \neq k}} \left( \frac{\mu_0}{2\pi} \ln \frac{1}{d_{kk'}} \right) i_{k'}$ 

mutual inductances henrys/m

#### 1. Series inductance *l*

• total flux linkages  $\lambda_k$  of conductor k depends on all currents  $i_{k'}$ 

$$\lambda_k = \left(\frac{\mu_0}{2\pi} \ln \frac{1}{r'_k}\right) i_k +$$

• in vector form:  $\lambda = Li$ 

• Faraday's law: 
$$v(t) = \frac{d}{dt}\lambda(t)$$

#### voltage drop along conductor

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$$\sum_{\substack{k'\neq k}} \left( \frac{\mu_0}{2\pi} \ln \frac{1}{d_{kk'}} \right) i_{k'}$$

$$= L \frac{d}{dt}i(t)$$

- 2. Shunt capacitance *c* 
  - voltage on surface of conductor k relative to reference:

$$v_k = \left(\frac{1}{2\pi\epsilon}\ln\frac{1}{r_k}\right)q_k +$$

- in vector form: v = Fq

$$\sum_{\substack{k'\neq k}} \left( \frac{1}{2\pi\epsilon} \ln \frac{1}{d_{kk'}} \right) q_{k'}$$

• let  $C := F^{-1}$ .  $C_{kk}$ : self capacitance/m,  $C_{kk'}$ : mutual capacitance/m

- 2. Shunt capacitance *c* 
  - voltage on surface of conductor k relative to reference:

$$v_k = \left(\frac{1}{2\pi\epsilon}\ln\frac{1}{r_k}\right)q_k +$$

• in vector form: v = Fq

$$\frac{d}{dt}v(t) = Fi(t)$$



## **Balanced 3***\phi* line

#### **Assumptions:**

- 1. Conductors equally spaced at D with equal radii r
- 2.  $i_1(t) + \cdots + i_n(t) = 0$  for all t
- 3.  $q_1(t) + \dots + q_n(t) = 0$  for all t





#### Phases are decoupled







#### Phases are decoupled



$$v_k = \left(\frac{1}{2\pi\epsilon} \ln \frac{D}{r}\right) q_k$$

 $(\cdot)^{-1}$ : capacitance c (F/m)



## Outline

#### 1. Line characteristics

- 2. Line models
  - Transmission matrix
  - In circuit model
  - Real and reactive line losses
  - Lossless line
  - Short line

## **Balanced 3***\phi* line

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#### **Per-phase line characteristics:**

series impedance / meter

 $z := r + i\omega l$   $\Omega/m$  $\Omega^{-1}/m$ shunt admittance / meter to neutral  $y := g + i\omega c$ 

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#### **Transmission matrix**



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ODE:  $\begin{bmatrix} \frac{dV}{dx} \\ \frac{dI}{dI} \end{bmatrix} = \begin{bmatrix} 0 & z \\ y & 0 \end{bmatrix} \begin{bmatrix} V \\ I \end{bmatrix}$ dx

boundary cond:

 $V(0) = V_2, I(0) = I_2$ 



#### **Transmission matrix**



 $m^{-1}$ characteristic impedance  $Z_c := \sqrt{\frac{z}{y}} \Omega/m$ propagation constant  $\gamma := \sqrt{zy}$ 

$$\begin{bmatrix} V(x) \\ I(x) \end{bmatrix} = U \begin{bmatrix} e^{\gamma x} & 0 \\ 0 & e^{-\gamma x} \end{bmatrix} U^{-1} \begin{bmatrix} V_2 \\ I_2 \end{bmatrix}$$
$$U := \begin{bmatrix} Z_c & -Z_c \\ 1 & 1 \end{bmatrix}, \quad U^{-1} := \frac{1}{2Z_c} \begin{bmatrix} 1 \\ -1 \end{bmatrix}$$





### **Transmission matrix**

Transmission matrix maps receiving-end  $(V_2, I_2)$  to sending-end  $(V_1, I_1)$ 

$$\begin{bmatrix} V_1 \\ I_1 \end{bmatrix} = \begin{bmatrix} \cosh(\gamma \ell) \\ Z_c^{-1} \sinh(\gamma \ell) \end{bmatrix}$$

characteristic impedance  $Z_c$ 

propagation constant  $\gamma$ 

# $\begin{array}{c} \operatorname{rd} \left( V_{2}, I_{2} \right) \text{ to sending-end } \left( V_{1}, I_{1} \right) \\ Z_{c} \sinh(\gamma \ell) \\ \cosh(\gamma \ell) \end{array} \begin{bmatrix} V_{2} \\ I_{2} \end{bmatrix}$

$$:= \sqrt{\frac{z}{y}} \quad \Omega/m$$
$$:= \sqrt{zy} \quad m^{-1}$$

## **П** circuit model



$$Z' = Z_c \sinh(\gamma \ell) = \sqrt{\frac{z}{y}} \sinh(\gamma \ell) = Z \frac{\sinh(\gamma \ell)}{\gamma \ell}$$
$$\frac{Y'}{2} = \frac{1}{Z_c} \frac{\cosh(\gamma \ell) - 1}{\sinh(\gamma \ell)} = \frac{1}{Z_c} \frac{\sinh(\gamma \ell/2)}{\cosh(\gamma \ell/2)} = \frac{Y}{2} \frac{\tanh(\gamma \ell)}{Z_c}$$

**EE/CS/EST 135** Steven Low Caltech Kirchhoff's laws:

$$I_{1} = \frac{Y'}{2}V_{1} + \frac{Y'}{2}V_{2} + I_{2}$$
$$V_{1} - V_{2} = Z'\left(\frac{Y'}{2}V_{2} + I_{2}\right)$$

 $h(\gamma \ell/2)$  $\gamma \ell/2$ 

 $Z := z\ell, \ Y := y\ell$ 



## **П** circuit model

Long line ( $\ell > 150$  miles) :

Medium line (50 <  $\ell$  < 150 miles): use  $Z = z\ell$  and  $Y = i\omega C$ 

Short line ( $\ell < 50$  miles) :

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use Z' and Y'use  $Z = z\ell$  and  $Y = i\omega 0$ use  $Z = z\ell$  and Y = 0

# Line loss

#### Sending-end current

$$I_{12} = \frac{1}{Z'}(V_1 - V_2) + \frac{Y'}{2}V_1$$
$$I_{21} = \frac{1}{Z'}(V_2 - V_1) + \frac{Y'}{2}V_2$$

Real and reactive line losses

$$I_{12} + I_{21} = (Y'/2)^H \left( |V_1|^2 + |V_1|^2 \right)^H \left( |V_1|^2$$

If Y' = 0 then  $I_{12} = -I_{21}$  sending current = receiving current

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#### $(I_{12} = I_1, I_{21} = -I_2)$

 $V_2|^2$ )

# Line loss

Sending-end line power

$$S_{12} := V_1 I_{12}^H = \left(\frac{1}{Z'}\right)^H \left(|V_1|^2 - V_1 V_2^H\right) + \left(\frac{Y'}{2}\right)^H |V_1|^2$$
$$S_{21} := V_2 (I_{21})^H = \left(\frac{1}{Z'}\right)^H \left(|V_2|^2 - V_2 V_1^H\right) + \left(\frac{Y'}{2}\right)^H |V_2|^2$$

Real and reactive line losses

$$S_{12} + S_{21} = Z' |I_{12}^s|^2 + \left(\frac{Y'}{2}\right)$$

$${}^{H}\left( |V_1|^2 + |V_2|^2 \right)$$

## **Lossless line:** r = g = 0

Characteristic impedance is real

$$Z_c = \sqrt{\frac{z}{y}} = \sqrt{\frac{i\omega l}{i\omega c}}$$

Propagation constant is imaginary

$$\gamma = \sqrt{zy} = \sqrt{(i\omega l)(i\omega d)}$$

 $\Pi$  circuit model: both series impedance and shunt admittance are reactive:

$$Z' = i Z_c \sin(\beta \ell) \quad \Omega,$$

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 $= \sqrt{\frac{l}{c}} \Omega$  $\overline{pc} = i\omega\sqrt{lc} m^{-1}$ 

$$\frac{Y'}{2} = i \frac{\omega c\ell}{2} \frac{\tan(\beta \ell/2)}{\beta \ell/2} \Omega^{-1}$$

#### Lossless line: r = g = 0

Voltage along the line



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 $\beta := \omega \sqrt{lc}$ 

Generally voltage drops along the line towards load

## Short line: Y = 0

Sending-end power from *i* to *j*:

$$S_{ij} = V_i I_{ij}^* = V_i \frac{V_i^* - V_j^*}{Z^*}$$

 $\stackrel{*}{-} = \frac{1}{Z^*} \left( |V_i|^2 - V_i V_j^* \right)$ 

#### Short line: Y = 0Load voltage solution and collapse

Receiving-end load power at bus 2:

$$-S_{21} = -V_2 I_{21}^* = -\frac{1}{Z^*} \left( |V_2|^2 - V_2 V_1^* \right)$$

Load power:  $P + iQ := -S_{21}$ 

Express  $-S_{21}$  in terms of load power P to relate load voltage  $|V_2|$  to P

$$-S_{21} = P(1 + i \tan \phi)$$

 $\phi := \theta_{V_2} - \theta_{-I_{21}}$ : load power factor angle  $\mathbf{Z}\mathbf{I}$ 

#### Short line: Y = 0Load voltage solution and collapse

How does load voltage  $|V_2|$  depend on active load power P?  $P(1 + i \tan \phi) = -\frac{1}{Z^*} \left( |V_2| \right)$ 

Assume:  $V_1 := |V_1| \angle 0^\circ \Rightarrow \theta_{21} := \theta_2 - \theta_1 = \theta_2$ 

- 2 real equations in  $|V_2|$  and  $\theta_{21}$  with P as parameter
- Solve for load voltage  $|V_2|$  given any P
- As load power P increases, solutions  $|V_2|$  trace out a nose curve
- If P increases further, no real solutions for  $|V_2|$  exists voltage collapse

$$V_2|^2 - |V_2||V_1|e^{i\theta_{21}}$$

## Short & lossless line: R = 0, Y = 0

Sending-end power from i to j:

$$S_{ij} = \frac{i}{X} \left( |V_i|^2 - V_i V_j^* \right)$$

Hence

$$P_{12} = \frac{|V_1| |V_2|}{X} \sin \theta_{12}$$
$$Q_{12} = \frac{1}{X} \left( |V_1|^2 - |V_1| |V_2| \cos \theta_{12} \right)$$
$$Q_{21} = \frac{1}{X} \left( |V_2|^2 - |V_1| |V_2| \cos \theta_{12} \right)$$

 $\operatorname{os} \theta_{12}$ 

bs  $\theta_{12}$ 

# Short & lossless line: R = 0, Y = 0

1. DC power flow model: R = 0, fixed  $|V_i|$ ,  $\sin \theta_{12} \approx \theta_{12}$ , ignore  $Q_{ij}$  $P_{ij} = \frac{|V_1| |V_2|}{X} \theta_{12}$ 

2. Decoupling:  $\frac{\partial P_{12}}{\partial |V_i|} = \frac{|V_j|}{X} \sin \theta_{12} \approx 0$   $\frac{\partial P_{12}}{\partial \theta_{12}} = \frac{|V_1| |V_2|}{X} \cos \theta_{12}$ 

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 $\frac{\partial P_{12}}{\partial |V_i|} = \frac{|V_j|}{X} \sin \theta_{12} \approx 0 \qquad \frac{\partial Q_{ij}}{\partial \theta_{12}} = \frac{|V_1| |V_2|}{X} \sin \theta_{12} \approx 0$ 

 $\cos \theta_{12} \approx \frac{|V_1||V_2|}{X}$ 

# Short & lossless line: R = 0, Y = 0

#### 3. Voltage regulation

$$\frac{\partial Q_{12}}{\partial |V_2|} = -\frac{|V_1|}{X} \cos \theta_{12} < 0$$
  
$$\frac{\partial Q_{21}}{\partial |V_2|} = \frac{1}{X} (2|V_2| - |V_1| \cos \theta_{12})$$

Voltage regulation: maintain high  $|V_2|$ :

- decrease  $Q_{12}$
- increase  $Q_{21}$

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 $(\theta_{12}) > 0$