

Power System Analysis

Chapter 4 Bus injection models

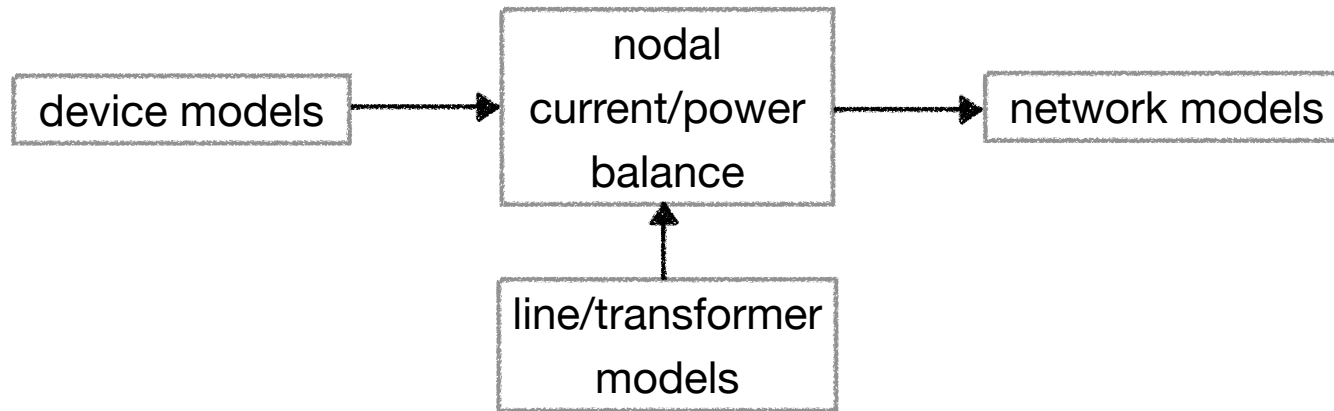
Outline

1. Component models
2. Network model: IV relation
3. Network model: sV relation
4. Computation methods
5. Linear power flow model

Outline

1. Component models
 - Sources, impedance
 - Transmission or distribution line
 - Transformer
2. Network model: IV relation
3. Network model: sV relation
4. Computation methods
5. Linear power flow model

Overview



single-phase or 3-phase

Single-phase devices

1. Single-terminal device j

- Voltage source (E_j, z_j) , current source (J_j, y_j) , power source (σ_j, z_j) , impedance z_j
- Terminal variables (V_j, I_j, s_j)
- External model: relation between (V_j, I_j) or (V_j, s_j)

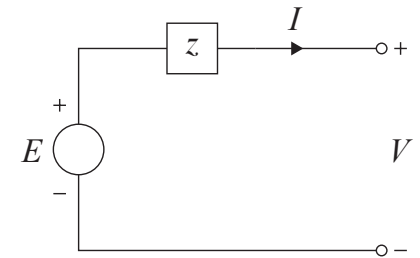
2. Two-terminal device (j, k)

- Line $(y_{jk}^s, y_{jk}^m, y_{kj}^m)$, transformer $(K_{jk}(n), \tilde{y}_{jk}^s, \tilde{y}_{jk}^m)$
- Terminal variables (V_j, I_{jk}, S_{jk}) and (V_k, I_{kj}, S_{kj})
- External model: relation between $(V_j, V_k, I_{jk}, I_{kj})$ or $(V_j, V_k, S_{jk}, S_{kj})$

Single-phase devices

1. Voltage source (E_j, z_j)

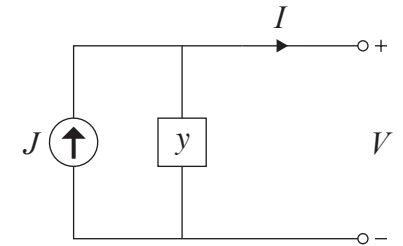
- Constant internal voltage E_j with series impedance z_j
- Models for Thevenin equivalent circuit of a balanced synchronous machine, secondary side of transformer, grid-forming inverter
- External model: $V_j = E_j - z_j I_j$
- External model: $s_j = V_j I_j^H = y_j^H V_j (E_j - V_j)^H$



Single-phase devices

2. Current source (J_j, y_j)

- Constant internal current J_j with shunt admittance y_j
- Models for Norton equivalent circuit of a synchronous generator, load (e.g. electric vehicle charger), grid-following inverter
- External model: $I_j = J_j - y_j V_j$
- External model: $s_j = V_j I_j^H = V_j (J_j - y_j V_j)^H$



Single-phase devices

3. Power source (σ_j, z_j)

- Constant internal power σ_j in series with impedance z_j
- Models for load, generator, secondary side of transformer
- External model: $\sigma_j = (V_j - z_j I_j) I_j^H$
- External model: $s_j = V_j I_j^H = \sigma_j + z_j I_j I_j^H$

Single-phase devices

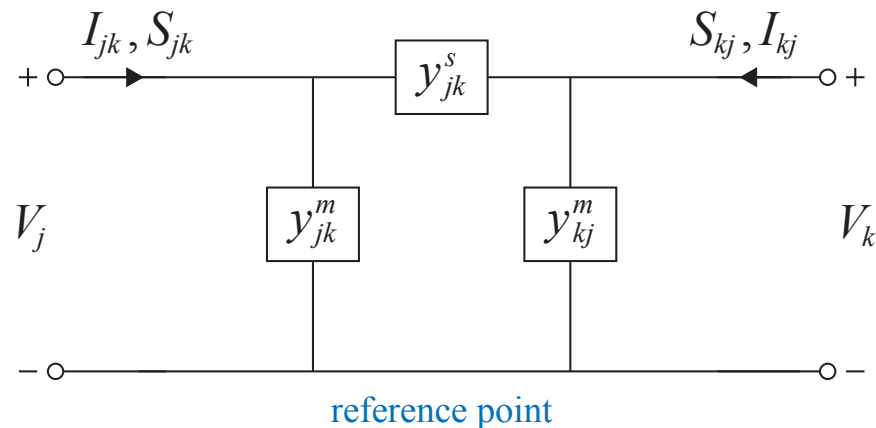
4. Impedance z_j

- Constant impedance z
- Models for load
- External model: $V_j = z_j I_j$

- External model: $s_j = V_j I_j^H = \frac{|V_j|^2}{z_j^H}$

Single-phase line $(y_{jk}^s, y_{jk}^m, y_{kj}^m)$

VI relation: Π circuit and admittance matrix Y_{line}



$$\begin{bmatrix} I_{jk} \\ I_{kj} \end{bmatrix} = \underbrace{\begin{bmatrix} y_{jk}^s + y_{jk}^m & -y_{jk}^s \\ -y_{jk}^s & y_{jk}^s + y_{kj}^m \end{bmatrix}}_{Y_{\text{line}}} \begin{bmatrix} V_j \\ V_k \end{bmatrix}$$

$$I_{jk} = y_{jk}^s(V_j - V_k) + y_{jk}^m V_j,$$

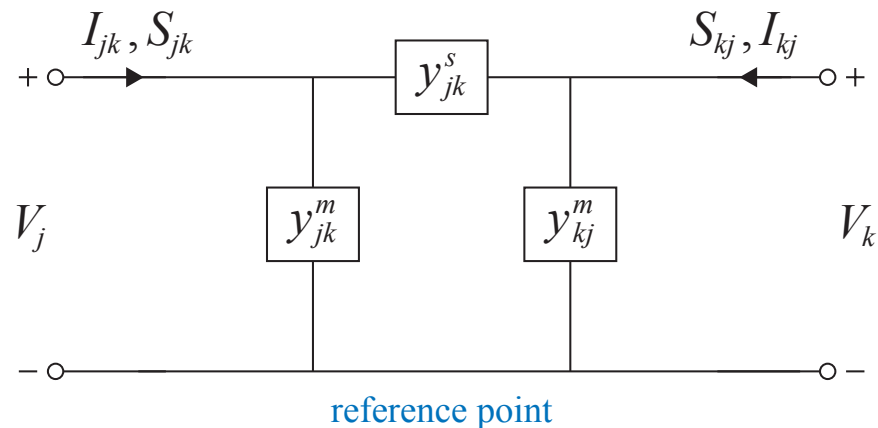
$$I_{kj} = y_{jk}^s(V_k - V_j) + y_{kj}^m V_k$$

admittance matrix Y_{line} :

- complex symmetric
- $[Y]_{jk} = -$ series admittance

Single-phase line $(y_{jk}^s, y_{jk}^m, y_{kj}^m)$

VI relation: Π circuit and admittance matrix Y_{line}



$$I_{jk} = y_{jk}^s(V_j - V_k) + y_{jk}^m V_j,$$

$$I_{kj} = y_{jk}^s(V_k - V_j) + y_{kj}^m V_k$$

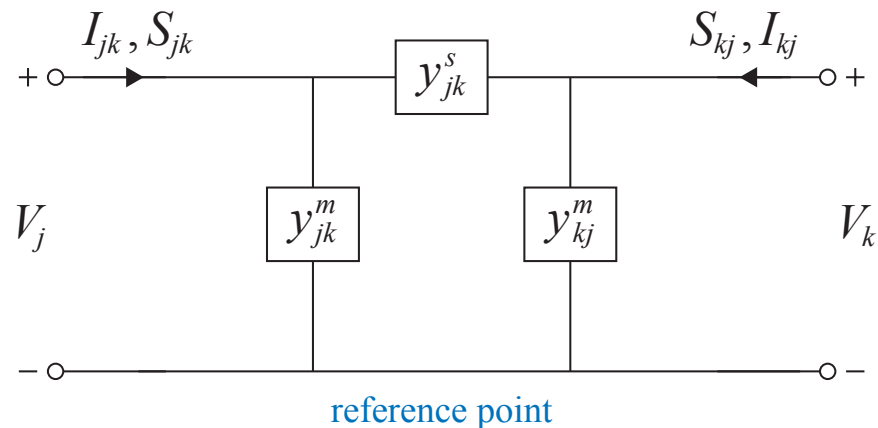
Their sum is total line current loss

$$I_{jk} + I_{kj} = y_{jk}^m V_j + y_{kj}^m V_k \neq 0$$

If $y_{jk}^m = y_{kj}^m = 0$, then $I_{jk} = -I_{kj}$

Single-phase line $(y_{jk}^s, y_{jk}^m, y_{kj}^m)$

V_S relation



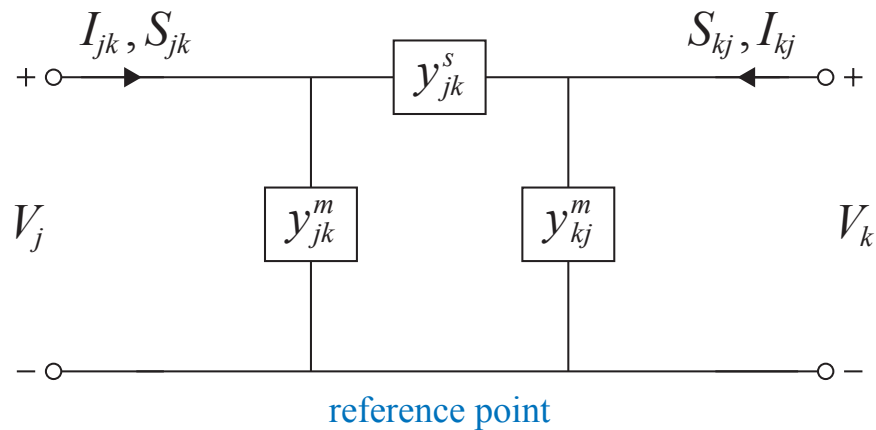
$$S_{jk} := V_j I_{jk}^H = \left(y_{jk}^s\right)^H \left(|V_j|^2 - V_j V_k^H\right) + \left(y_{jk}^m\right)^H |V_j|^2$$

$$S_{kj} := V_k I_{kj}^H = \left(y_{jk}^s\right)^H \left(|V_k|^2 - V_k V_j^H\right) + \left(y_{kj}^m\right)^H |V_k|^2$$

quadratic equations

Single-phase line $(y_{jk}^s, y_{jk}^m, y_{kj}^m)$

V_s relation

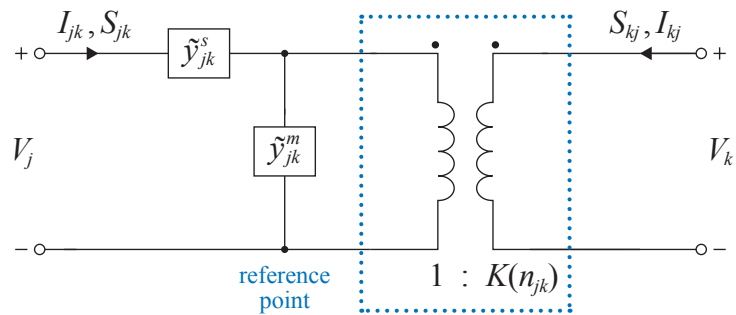


Line loss

$$S_{jk} + S_{kj} = \underbrace{\left(y_{jk}^s\right)^H \left|V_j - V_k\right|^2}_{\text{series loss}} + \underbrace{\left(y_{jk}^m\right)^H \left|V_j\right|^2 + \left(y_{kj}^m\right)^H \left|V_k\right|^2}_{\text{shunt loss}}$$

Single-phase transformer $(K(n_{jk}), \tilde{y}_{jk}^s, \tilde{y}_{jk}^m)$

Complex $K(n_{jk})$

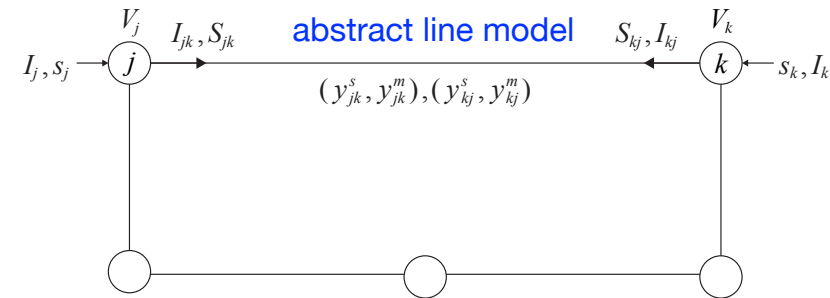
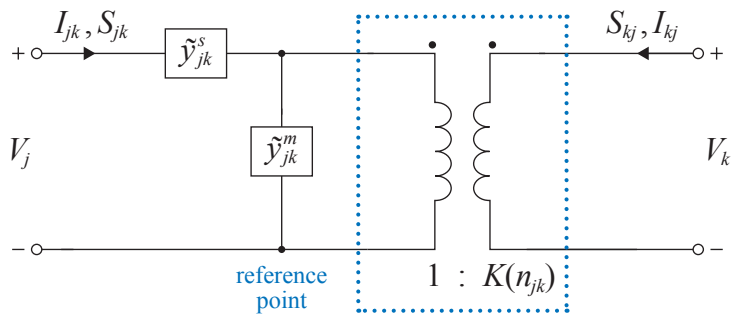


$$\begin{bmatrix} I_{jk} \\ I_{kj} \end{bmatrix} = \underbrace{\begin{bmatrix} y_{jk}^s & -y_{jk}^s / K_{jk}(n) \\ -y_{jk}^s / \bar{K}_{jk}(n) & (y_{jk}^s + y_{jk}^m) / |K_{jk}(n)|^2 \end{bmatrix}}_{Y_{\text{transformer}}} \begin{bmatrix} V_j \\ V_k \end{bmatrix}$$

- $Y_{\text{transformer}}$: *not* symmetric
- Has no equivalent Π circuit
- Use admittance or transmission matrix for analysis

Single-phase transformer $(K(n_{jk}), \tilde{y}_{jk}^s, \tilde{y}_{jk}^m)$

Complex $K(n_{jk})$

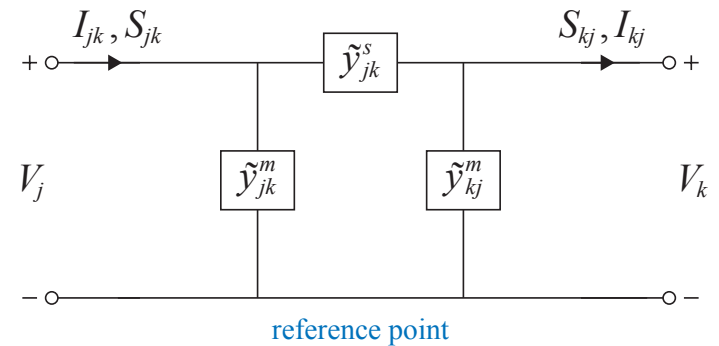
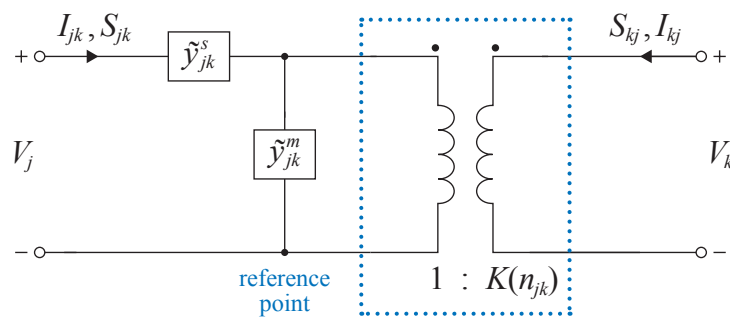


$$\begin{bmatrix} I_{jk} \\ I_{kj} \end{bmatrix} = \underbrace{\begin{bmatrix} y_{jk}^s & -y_{jk}^s / K_{jk}(n) \\ -y_{jk}^s / \bar{K}_{jk}(n) & (y_{jk}^s + y_{jk}^m) / |K_{jk}(n)|^2 \end{bmatrix}}_{Y_{\text{transformer}}} \begin{bmatrix} V_j \\ V_k \end{bmatrix}$$

$$\begin{aligned} y_{jk}^s &:= \frac{\tilde{y}_{jk}^s}{K_{jk}(n)}, & y_{jk}^m &:= \left(1 - \frac{1}{K_{jk}(n)}\right) \tilde{y}_{jk}^s \\ y_{kj}^s &:= \frac{\tilde{y}_{jk}^s}{\bar{K}_{jk}(n)}, & y_{kj}^m &:= \frac{1 - K_{jk}(n)}{|K_{jk}(n)|^2} \tilde{y}_{jk}^s + \frac{1}{|K_{jk}(n)|^2} \tilde{y}_{jk}^m \end{aligned}$$

Single-phase transformer $(K(n_{jk}), \tilde{y}_{jk}^s, \tilde{y}_{jk}^m)$

Real $K(n_{jk}) = n_{jk}$



$$I_{jk} = y_{jk}^s (V_j - a_{jk} V_k)$$

$$I_{jk} = y_{jk}^m a_{jk} V_k + n_{jk} (-I_{kj})$$

$$y_{jk}^s := a_{jk} \tilde{y}_{jk}^s = y_{kj}^s$$

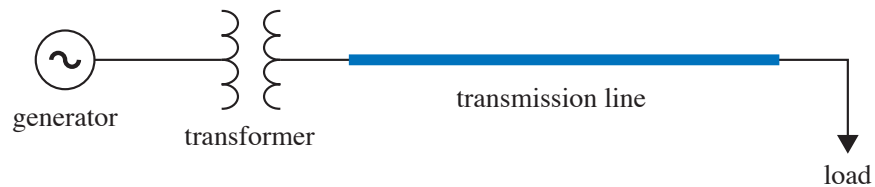
$$y_{jk}^m := (1 - a_{jk}) \tilde{y}_{jk}^s \quad \tilde{y}_{jk}^m \neq \tilde{y}_{kj}^m$$

$$y_{kj}^m := a_{jk} (a_{jk} - 1) \tilde{y}_{jk}^s + a_{jk}^2 \tilde{y}_{jk}^m$$

Outline

1. Component models
2. Network model: IV relation
 - Example and network model
 - Admittance matrix Y and properties
 - Kron reduction Y/Y_{22} and properties
 - Radial network
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Example



System

- Generator: current source (I_1, y_1)
- Transformer $(n, \tilde{y}^s, \tilde{y}^m)$
- Transmission line with series admittance y
- Load: current source (I_2, y_2)

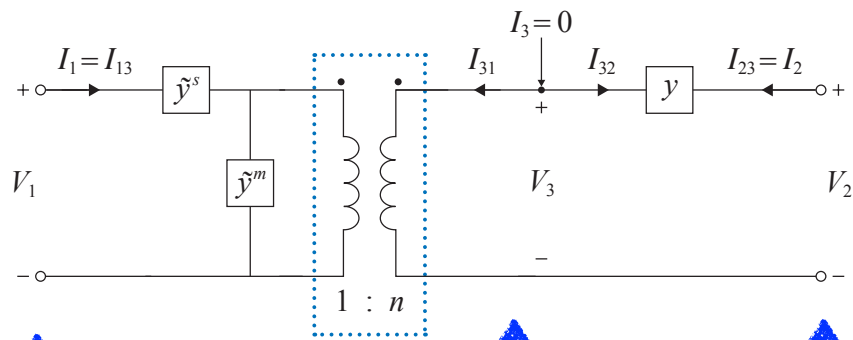
Derive

- Derive network model (admittance matrix Y)

Derive Y in 2 steps

Example

Step 1: transformer + line



$$\begin{bmatrix} I_{13} \\ I_{31} \end{bmatrix} = \begin{bmatrix} \tilde{y}^s & -a\tilde{y}^s \\ -a\tilde{y}^s & a^2(\tilde{y}^s + \tilde{y}^m) \end{bmatrix} \begin{bmatrix} V_1 \\ V_3 \end{bmatrix}$$

relate branch currents with nodal voltages

$$\begin{bmatrix} I_{32} \\ I_{23} \end{bmatrix} = \begin{bmatrix} y & -y \\ -y & y \end{bmatrix} \begin{bmatrix} V_3 \\ V_2 \end{bmatrix}$$

Nodal current balance (KCL):

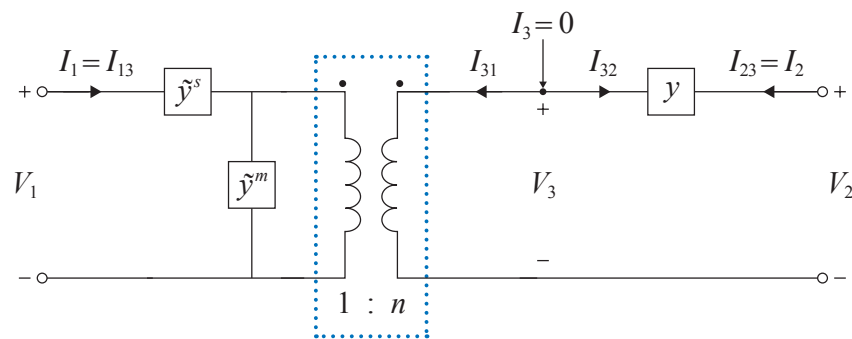
$$I_1 = I_{13}$$

$$I_3 = I_{31} + I_{32} = 0$$

$$I_2 = I_{23}$$

Example

Step 1: transformer + line



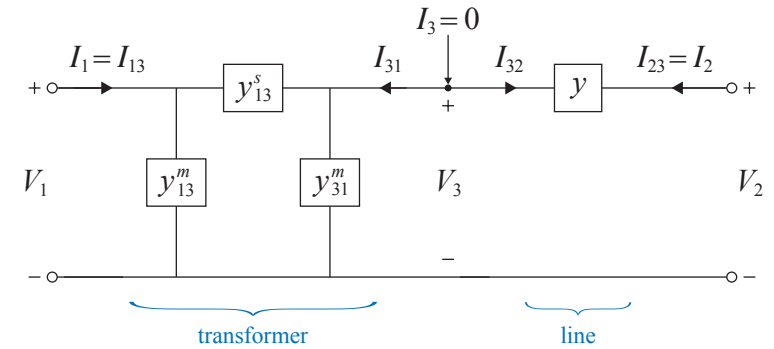
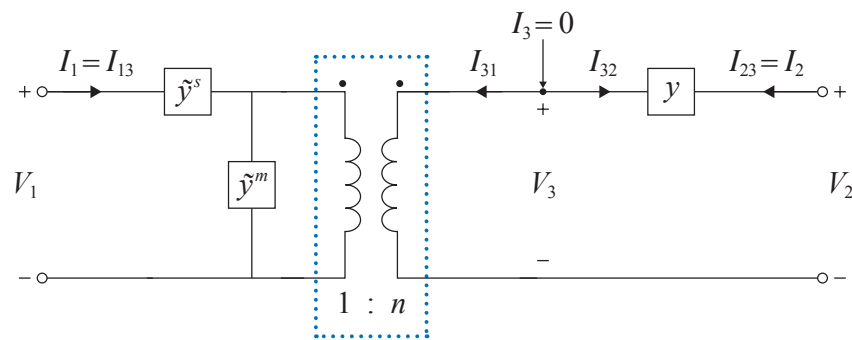
Eliminate branch currents:

$$\begin{bmatrix} I_1 \\ I_2 \\ I_3 \end{bmatrix} = \underbrace{\begin{bmatrix} \tilde{y}^s & 0 & -a\tilde{y}^s \\ 0 & y & -y \\ -a\tilde{y}^s & -y & y + a^2(\tilde{y}^s + \tilde{y}^m) \end{bmatrix}}_{Y_1} \begin{bmatrix} V_1 \\ V_2 \\ V_3 \end{bmatrix}$$

- Y_1 : complex symmetric
- Hence: admittance matrix with Π circuit
- Unequal shunt elements (even if $\tilde{y}^m = 0$)

Example

Step 1: transformer + line



Eliminate branch currents:

$$\begin{bmatrix} I_1 \\ I_2 \\ I_3 \end{bmatrix} = \underbrace{\begin{bmatrix} \tilde{y}^s & 0 & -a\tilde{y}^s \\ 0 & y & -y \\ -a\tilde{y}^s & -y & y + a^2(\tilde{y}^s + \tilde{y}^m) \end{bmatrix}}_{Y_1} \begin{bmatrix} V_1 \\ V_2 \\ V_3 \end{bmatrix}$$

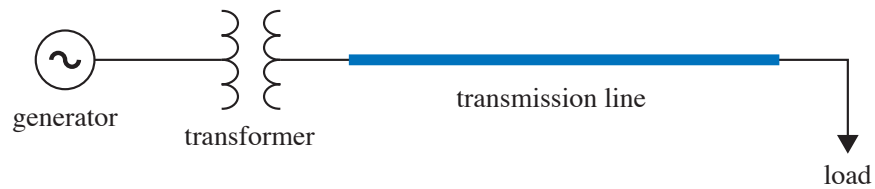
$$y_{13}^s := a\tilde{y}^s$$

$$y_{13}^m := (1-a)\tilde{y}^s$$

$$y_{31}^m := a(a-1)\tilde{y}^s + a^2\tilde{y}^m$$

Example

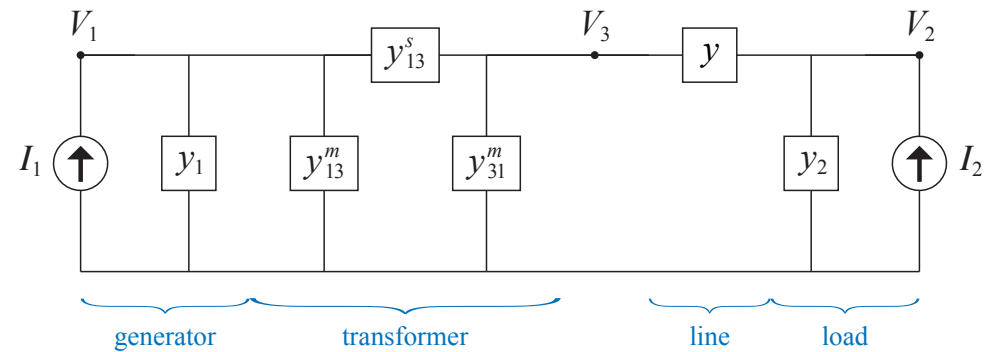
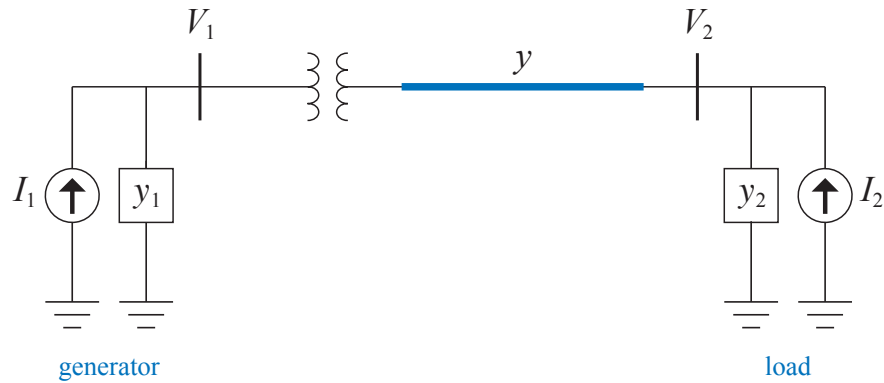
Step 2: overall system



generator/load admittances

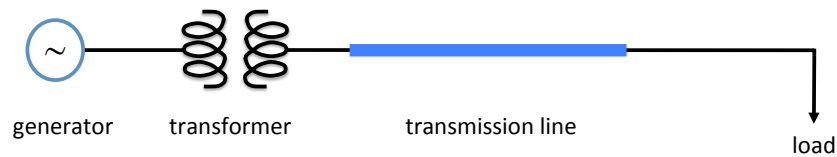
$$\begin{bmatrix} I_1 \\ I_2 \\ 0 \end{bmatrix} = \begin{bmatrix} \tilde{y}^s + y_1 & 0 & -a\tilde{y}^s \\ 0 & y + y_2 & -y \\ -a\tilde{y}^s & -y & y + a^2(\tilde{y}^s + \tilde{y}^m) \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \\ V_3 \end{bmatrix}$$

Red arrows point from the text "generator/load admittances" to the terms y_1 and y_2 in the matrix, which are enclosed in red boxes.



Example

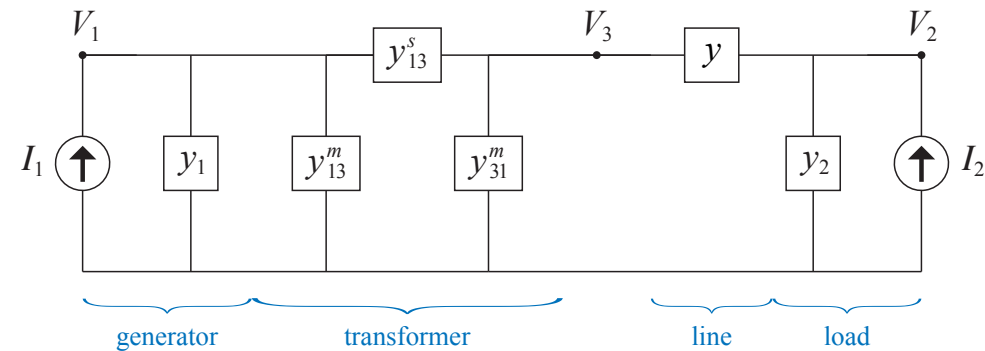
Step 2: overall system



generator/load admittances

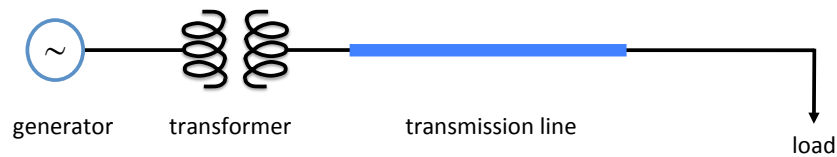
$$\begin{bmatrix} I_1 \\ I_2 \\ 0 \end{bmatrix} = \begin{bmatrix} \tilde{y}^s + y_1 & 0 & -a\tilde{y}^s \\ 0 & y + y_2 & -y \\ -a\tilde{y}^s & -y & y + a^2(\tilde{y}^s + \tilde{y}^m) \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \\ V_3 \end{bmatrix}$$

- Overall network model: **ideal** current sources connected by network
- Network: admittance matrix Y
- Y includes admittances of non-ideal current sources



Example

Step 2: overall system



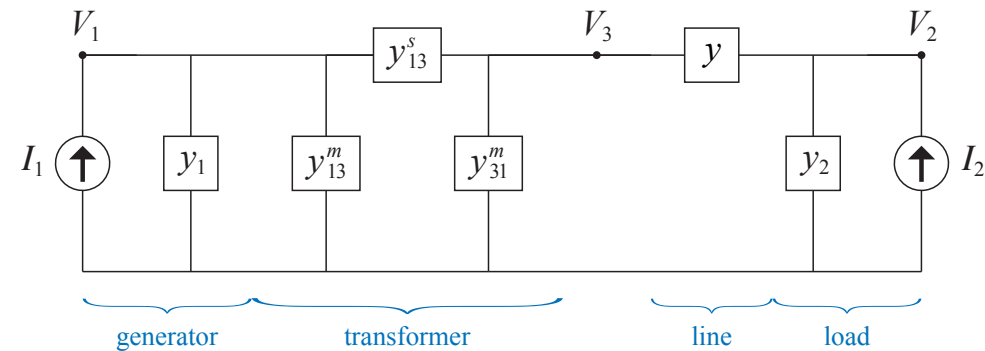
Kron reduction (see below)

- Internal bus has zero injection $I_3 = 0$
- Can eliminate (V_3, I_3)
- External behavior: relation between (I_1, I_2) and (V_1, V_2)

generator/load admittances

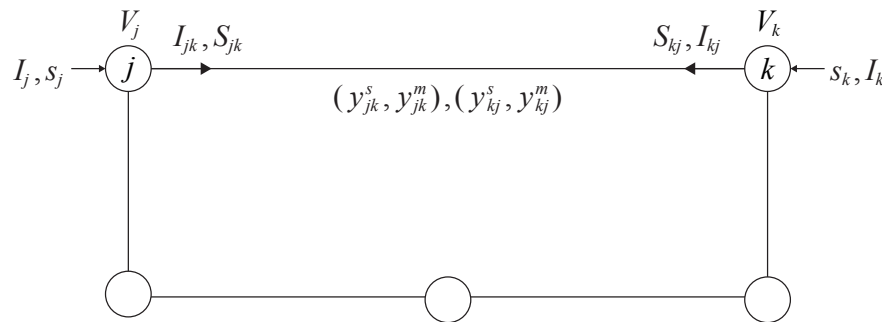
$$\begin{bmatrix} I_1 \\ I_2 \\ 0 \end{bmatrix} = \begin{bmatrix} \tilde{y}^s + y_1 & 0 & -a\tilde{y}^s \\ 0 & y + y_2 & -y \\ -a\tilde{y}^s & -y & y + a^2(\tilde{y}^s + \tilde{y}^m) \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \\ V_3 \end{bmatrix}$$

The equation shows the relationship between currents I_1, I_2, I_3 and voltages V_1, V_2, V_3 . The matrix elements are admittances. Red boxes highlight y_1 and y_2 in the original image, with red arrows pointing to the text "generator/load admittances".

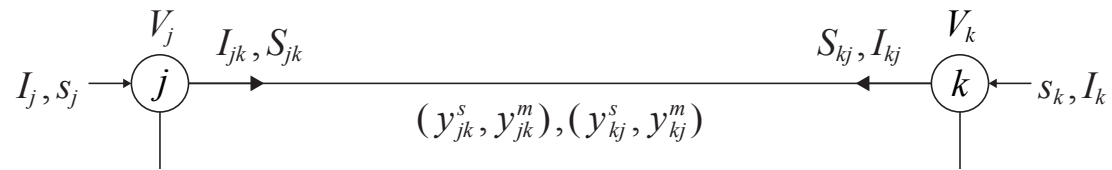


Line model

- Network $G := (\bar{N}, E)$
 - $\bar{N} := \{0\} \cup N := \{0\} \cup \{1, \dots, N\}$: buses/nodes/terminals
 - $E \subseteq \bar{N} \times \bar{N}$: lines/branches/links/edges
- Each line (j, k) is parameterized by (y_{jk}^s, y_{jk}^m) and (y_{kj}^s, y_{kj}^m)
 - (y_{jk}^s, y_{jk}^m) : series and shunt admittances from j to k
 - (y_{kj}^s, y_{kj}^m) : series and shunt admittances from k to j
 - Models transmission or distribution lines, single-phase transformers



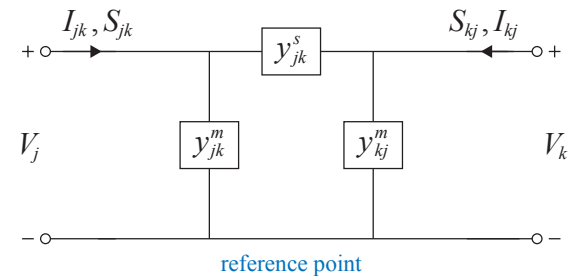
Line model



Sending-end currents

$$I_{jk} = y_{jk}^s(V_j - V_k) + y_{jk}^m V_j, \quad I_{kj} = y_{kj}^s(V_k - V_j) + y_{kj}^m V_k,$$

If $y_{jk}^s = y_{kj}^s$: same relation but equivalent to Π circuit:



Network model

Nodal current balance

$$I_j = \sum_{k:j \sim k} I_{jk}$$

Network model

Nodal current balance

$$I_j = \sum_{k:j \sim k} I_{jk} = \left(\sum_{k:j \sim k} y_{jk}^s + y_{jj}^m \right) V_j - \sum_{k:j \sim k} y_{jk}^s V_k$$

total shunt admittance: $y_{jj}^m := \sum_{k:j \sim k} y_{jk}^m$

Network model

Admittance matrix Y

$$I_j = \sum_{k:j \sim k} I_{jk} = \left(\sum_{k:j \sim k} y_{jk}^s + y_{jj}^m \right) V_j - \sum_{k:j \sim k} y_{jk}^s V_k$$

In vector form:

$$I = YV \text{ where } Y_{jk} = \begin{cases} -y_{jk}^s, & j \sim k \ (j \neq k) \\ \sum_{l:j \sim l} y_{jl}^s + y_{jj}^m, & j = k \\ 0 & \text{otherwise} \end{cases}$$

Network model

Admittance matrix Y

Y can be written down by inspection of network graph

- Off-diagonal entry: $-$ series admittance
- Diagonal entry: \sum series admittances $+$ total shunt admittance

In vector form:

$$I = YV \text{ where } Y_{jk} = \begin{cases} -y_{jk}^s, & j \sim k \ (j \neq k) \\ \sum_{l:j \sim l} y_{jl}^s + y_{jj}^m, & j = k \\ 0 & \text{otherwise} \end{cases}$$

Network model

Admittance matrix Y

A matrix Y has a Π circuit representation

- if it is complex symmetric ($y_{jk}^s = y_{kj}^s$)

In vector form:

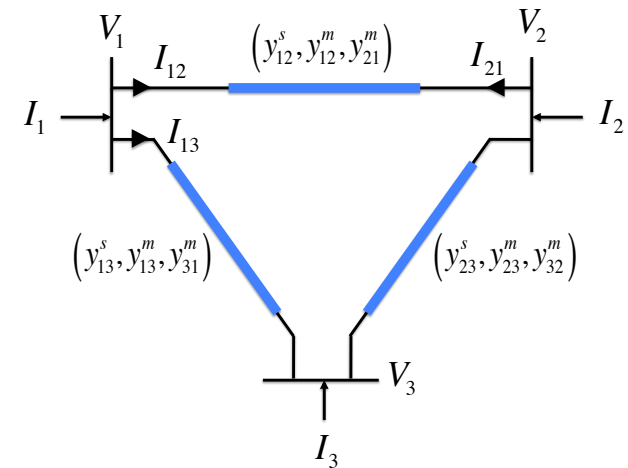
$$I = YV \text{ where } Y_{jk} = \begin{cases} -y_{jk}^s, & j \sim k \ (j \neq k) \\ \sum_{l:j \sim l} y_{jl}^s + y_{jj}^m, & j = k \\ 0 & \text{otherwise} \end{cases}$$

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Admittance matrix Y

Example



$$\begin{bmatrix} I_1 \\ I_2 \\ I_3 \end{bmatrix} = \begin{bmatrix} y_{12}^s + y_{13}^s + y_{11}^m & -y_{12}^s & -y_{13}^s \\ -y_{12}^s & y_{12}^s + y_{23}^s + y_{22}^m & -y_{23}^s \\ -y_{13}^s & -y_{23}^s & y_{13}^s + y_{23}^s + y_{33}^m \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \\ V_3 \end{bmatrix}$$

total shunt admittance: $y_{jj}^m := \sum_{k:j \sim k} y_{jk}^m$

Admittance matrix Y

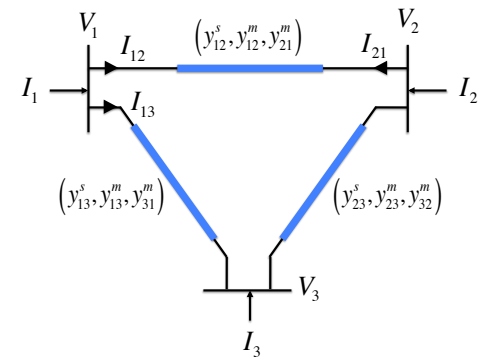
In terms of incidence matrix C

bus-by-line incidence matrix

$$C_{jl} = \begin{cases} 1 & \text{if } l = j \rightarrow k \text{ for some bus } k \\ -1 & \text{if } l = i \rightarrow j \text{ for some bus } i \\ 0 & \text{otherwise} \end{cases}$$

example:

$$C = \begin{bmatrix} 1 & 0 & -1 \\ -1 & 1 & 0 \\ 0 & -1 & 1 \end{bmatrix}$$



Admittance matrix Y

In terms of incidence matrix C

bus-by-line incidence matrix

$$C_{jl} = \begin{cases} 1 & \text{if } l = j \rightarrow k \text{ for some bus } k \\ -1 & \text{if } l = i \rightarrow j \text{ for some bus } i \\ 0 & \text{otherwise} \end{cases}$$

$$Y = CD_y^s C^\top + D_y^m$$

where $D_y^s := \text{diag}(y_l^s, l \in E)$, $D_y^m := \text{diag}(y_{jj}^m, j \in \bar{N})$

Y is a complex Laplacian matrix when $Y^m = 0$

Properties of Y

1. The inverse $Z := Y^{-1}$, if exists, is called a **bus impedance matrix** or an **impedance matrix**
 - Useful for fault analysis
 - Solving $I = YV$ for V
 - Advantages of Y : Y can be constructed by inspection of one-line diagram and inherits sparsity structure of G . Z can/does not.
2. Next: study existence of Z
 - Derive (Schur complement) expressions for Z , when Y is nonsingular
 - 4 sufficient conditions for Y to be nonsingular based on the expressions for Z

Inverse of Y

If exists

Let $Y := G + iB$, $Z := R + iX$

Y nonsingular $\iff \exists(R, X)$ s.t. $YZ = ZY = \mathbb{I}$

$\iff YZ = (GR - BX) + i(GX + BR) = \mathbb{I}$

$$\iff \underbrace{\begin{bmatrix} G & -B \\ B & G \end{bmatrix}}_M \begin{bmatrix} R \\ X \end{bmatrix} = \begin{bmatrix} I \\ 0 \end{bmatrix}$$

Suppose G is nonsingular. Then Y nonsingular \iff Schur complement $M/G := G + BG^{-1}B$ nonsingular

$$\text{Then } M^{-1} = \begin{bmatrix} (M/G)^{-1} & (M/G)^{-1}BG^{-1} \\ -G^{-1}B(M/G)^{-1} & G^{-1} - G^{-1}B(M/G)^{-1}BG^{-1} \end{bmatrix} \text{ and hence } \begin{bmatrix} R \\ X \end{bmatrix} = \begin{bmatrix} (M/G)^{-1} \\ -G^{-1}B(M/G)^{-1} \end{bmatrix}$$

Invertibility of Y

Theorem 1

Suppose Y is complex symmetric ($y_{jk}^s = y_{kj}^s$).

If $\text{Re}(Y) \succ 0$, then Y^{-1} exists, is symmetric, and $\text{Re}(Y^{-1}) \succ 0$

Proof

Let $Y = G + iB$ with $G \succ 0$. Then $M/G := G + BG^{-1}B \succ 0$ because $G, G^{-1} \succ 0$ and $B = B^T$.

Therefore both G and M/G are nonsingular, which implies that Y is nonsingular (from previous slide).

Moreover $\begin{bmatrix} R \\ X \end{bmatrix} = \begin{bmatrix} (M/G)^{-1} \\ -G^{-1}B(M/G)^{-1} \end{bmatrix}$ implies $\text{Re}(Y^{-1}) = (M/G)^{-1} \succ 0$ since $M/G \succ 0$.

Finally, to prove $Z := Y^{-1}$ is symmetric: substitute $Z^T Y^T = Z^T Y$ and $Y^T Z^T = YZ^T$ into (transpose of) $ZY = YZ = \mathbb{I}$ to get:

$$Z^T Y = Z^T Y^T = Y^T Z^T = YZ^T = \mathbb{I} \quad \text{i.e.,} \quad Z^T = Y^{-1} = Z$$

Invertibility of Y

Let $y_{jk}^s =: g_{jk}^s + ib_{jk}^s$, $y_{jk}^m =: g_{jk}^m + ib_{jk}^m$, $y_{kj}^m =: g_{kj}^m + ib_{kj}^m$

Conditions

1. $g_{jk}^s, g_{jk}^m, g_{kj}^m \geq 0$ for all lines $(j, k) \in E$, i.e., nonnegative conductances
2. $\sum_{k:k \sim j} g_{jk}^m \neq 0$ for all buses $j \in \bar{N}$, i.e., there is a shunt conductance incident on every bus
3. $g_{jk}^s \neq 0$ for all lines $(j, k) \in E$, and $\exists (j', k') \in E$ s.t. $g_{j'k'}^m \neq 0$, i.e., all series conductances are nonzero and there is at least one nonzero shunt conductance

Theorem 2

Suppose G is connected and Y is complex symmetric ($y_{jk}^s = y_{kj}^s$). If conditions 1 and either 2 or 3 are satisfied, then

1. $\text{Re}(Y) \succ 0$
2. Y^{-1} exists, is symmetric, and $\text{Re}(Y^{-1}) \succ 0$

Invertibility of Y

Theorem 2

Suppose G is connected and Y is complex symmetric ($y_{jk}^s = y_{kj}^s$). If conditions 1 and either 2 or 3 are satisfied, then

1. $\operatorname{Re}(Y) \succ 0$
2. Y^{-1} exists, is symmetric, and $\operatorname{Re}(Y^{-1}) \succ 0$

Proof

For any nonzero $\rho \in \mathbb{R}^{N+1}$, these conditions imply

$$\begin{aligned}\rho^\top G \rho &= \sum_j \sum_k \rho_j \rho_k G_{jk} = \sum_j \left(\sum_{k:j \sim k} -\rho_j \rho_k g_{jk}^s + \rho_j^2 \sum_{i:j \sim i} (g_{ji}^s + g_{ji}^m) \right) \\ &= \sum_{(j,k) \in E} \left(\rho_j^2 - 2\rho_j \rho_k + \rho_k^2 \right) g_{jk}^s + \sum_{j \in \bar{N}} \rho_j^2 \sum_{i:j \sim i} g_{ji}^m \\ &= \sum_{(j,k) \in E} \left(\rho_j - \rho_k \right)^2 g_{jk}^s + \sum_{j \in \bar{N}} \rho_j^2 \sum_{i:j \sim i} g_{ji}^m > 0\end{aligned}$$

Inverse of Y

If exists

Let $Y := G + iB$, $Z := R + iX$

$$Y \text{ nonsingular} \iff \underbrace{\begin{bmatrix} G & -B \\ B & G \end{bmatrix}}_M \begin{bmatrix} R \\ X \end{bmatrix} = \begin{bmatrix} I \\ 0 \end{bmatrix} \text{ which is the same as: } \underbrace{\begin{bmatrix} B & G \\ G & -B \end{bmatrix}}_{M'} \begin{bmatrix} R \\ X \end{bmatrix} = \begin{bmatrix} 0 \\ I \end{bmatrix}$$

Suppose B is nonsingular. Then Y nonsingular \iff Schur complement $M'/B := -(B + GB^{-1}G)$ nonsingular

$$\text{Then } M'^{-1} = \begin{bmatrix} B^{-1} + B^{-1}G(M'/B)^{-1}GB^{-1} & -B^{-1}G(M'/B)^{-1} \\ -(M'/B)^{-1}GB^{-1} & (M'/B)^{-1} \end{bmatrix} \text{ and hence } \begin{bmatrix} R \\ X \end{bmatrix} = \begin{bmatrix} -B^{-1}G(M'/B)^{-1} \\ (M'/B)^{-1} \end{bmatrix}$$

This leads to 2 analogous sufficient conditions in terms of $\text{Im}(Y)$ and $(b_{jk}^s, b_{jk}^m, b_{kj}^m)$ with similar proofs.

Invertibility of Y

Theorem 3

Suppose Y is complex symmetric ($y_{jk}^s = y_{kj}^s$).

If $\text{Im}(Y) < 0$, then Y^{-1} exists, is symmetric, and $\text{Im}(Y^{-1}) > 0$

Invertibility of Y

Let $y_{jk}^s =: g_{jk}^s + ib_{jk}^s$, $y_{jk}^m =: g_{jk}^m + ib_{jk}^m$, $y_{kj}^m =: g_{kj}^m + ib_{kj}^m$

Conditions

1. $b_{jk}^s, b_{jk}^m, b_{kj}^m \leq 0$ for all lines $(j, k) \in E$, i.e., nonpositive susceptances
2. $\sum_{k:k \sim j} b_{jk}^m \neq 0$ for all buses $j \in \bar{N}$, i.e., there is a shunt susceptance incident on every bus
3. $b_{jk}^s \neq 0$ for all lines $(j, k) \in E$, and $\exists (j', k') \in E$ s.t. $b_{j'k'}^m \neq 0$, i.e., all series susceptances are nonzero and there is at least one nonzero shunt susceptance

Theorem 4

Suppose G is connected and Y is complex symmetric ($y_{jk}^s = y_{kj}^s$). If conditions 1 and either 2 or 3 are satisfied, then

1. $\text{Im}(Y) < 0$
2. Y^{-1} exists, is symmetric, and $\text{Im}(Y^{-1}) > 0$

Invertibility of Y

Sufficiency only

These conditions on are sufficient only

- Conditions $(g_{jk}^s, g_{jk}^m, g_{kj}^m)$ in Theorem 2 are usually satisfied by transmission/distribution lines
- ... but not by transformers

Example:

Example 1 with node 3 at the **primary** side of the **ideal** transformer has an admittance matrix

$$Y = \begin{bmatrix} \tilde{y}^s & 0 & -\tilde{y}^s \\ 0 & y & -ny \\ -\tilde{y}^s & -ny & \tilde{y}^s + \tilde{y}^m + n^2y \end{bmatrix}$$

Suppose $g^s, \tilde{g}^s > 0$, $b^s, \tilde{b}^s \leq 0$, $\tilde{b}^m \geq 0$. Then $g_{23}^m := (1 - n)g^s$ and $g_{32}^m := n(n - 1)g^s$ have opposite signs ($n \neq 1$)

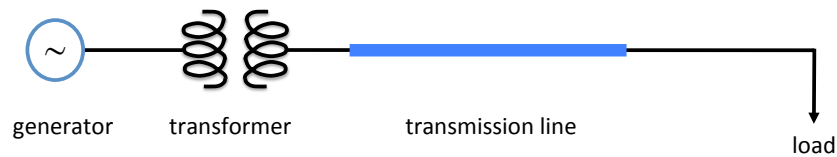
Hence Y does not satisfy conditions in Theorem 2. But Y is nonsingular if and only if $\tilde{b}_m > 0$

Outline

1. Component models
2. Network model: VI relation
 - Example and network model
 - Admittance matrix Y and properties
 - Kron reduction Y/Y_{22} and properties
 - Radial network
3. Network model: Vs relation
4. Computation methods
5. Linear power flow model

Example

Step 2: overall system

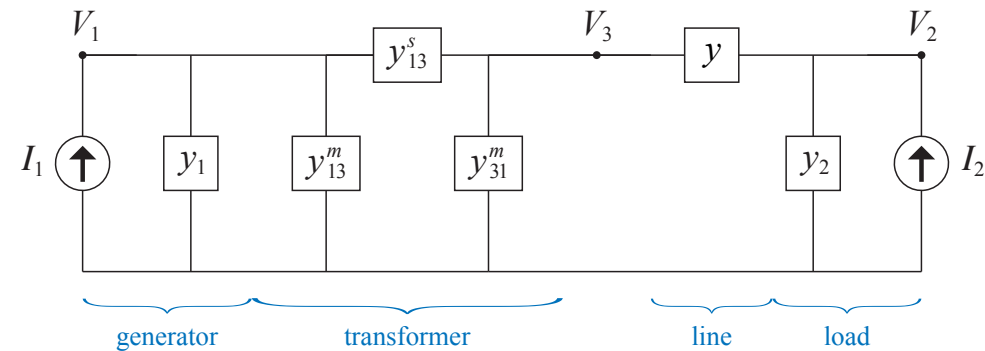


$$\begin{bmatrix} I_1 \\ I_2 \\ 0 \end{bmatrix} = \begin{bmatrix} \tilde{y}^s + y_1 & 0 & -a\tilde{y}^s \\ 0 & y + y_2 & -y \\ -a\tilde{y}^s & -y & y + a^2(\tilde{y}^s + \tilde{y}^m) \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \\ V_3 \end{bmatrix}$$

generator/load admittances

Kron reduction (see below)

- Internal bus has zero injection $I_3 = 0$
- Can eliminate (V_3, I_3)
- External behavior: relation between (I_1, I_2) and (V_1, V_2)



Kron reduction

- $N_{\text{red}} \subseteq \bar{N}$: buses of interest, e.g., terminal buses
- Want to relate current injections and voltages at buses in N_{red}

$$\begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \underbrace{\begin{bmatrix} Y_{11} & Y_{12} \\ Y_{21} & Y_{22} \end{bmatrix}}_Y \begin{bmatrix} V_1 \\ V_2 \end{bmatrix} \begin{array}{l} \leftarrow N_{\text{red}} \\ \leftarrow \bar{N} \setminus N_{\text{red}} \end{array}$$

- Eliminate $V_2 = -Y_{22}^{-1}Y_{21}V_1 + Y_{22}^{-1}I_2$
- giving $(Y_{11} - Y_{12}Y_{22}^{-1}Y_{21})V_1 = I_1 - Y_{12}Y_{22}^{-1}I_2$
Schur complement

Kron reduction

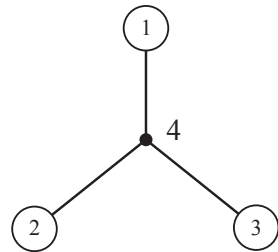
If internal injections $I_2 = 0$:

$$Y/Y_{22} := (Y_{11} - Y_{12}Y_{22}^{-1}Y_{21}) V_1 = I_1$$

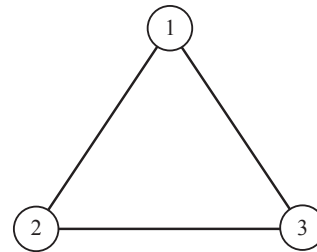
Schur complement

- Describes effective connectivity and line admittances of reduced network

Example:



(a) Original network



(b) Kron reduced network

Existence of Kron reduction

Admittance matrix $Y = CY^sC$ where $Y^s := \text{diag} \left(y_{jk}^s \right)$

When Y is **real**, it is called a real Laplacian matrix

- $(N + 1) \times (N + 1)$ real symmetric matrix
- Row sum = column sum = 0
- $\text{rank}(Y) = N$, $\text{null}(Y) = \text{span}(\mathbf{1})$ when all y_{jk}^s are (real &) of the **same** sign (otherwise $\text{rank}(Y)$ can be $< N$)
- Any principal submatrix is invertible, i.e., Y/Y_{22} always exists (we will study later in more detail for linear models)

When Y is a complex symmetric, but not Hermitian, these properties may not hold

In particular, Y_{22} may not be invertible and Y/Y_{22} may not exist

Existence of Kron reduction

Next: Properties of Y_{22} and Y/Y_{22}

- Conditions on $(y_{jk}^s, y_{jk}^m, y_{kj}^m)$ for Y_{22} to be nonsingular, hence existence of Y/Y_{22}
- Conditions on $(y_{jk}^s, y_{jk}^m, y_{kj}^m)$ for Y/Y_{22} to be nonsingular

Invertibility of Y_{22}

When $y_{jk}^s = y_{kj}^s$

Recall proof of Theorem 2:

$$\rho^\top G \rho = \sum_{(j,k) \in E} (\rho_j - \rho_k)^2 g_{jk}^s + \sum_{j \in \bar{N}} \rho_j^2 \sum_{i: j \sim i} g_{ji}^m > 0$$

↑
term associated with lines

↑
term associated with nodes

Invertibility of Y_{22}

When $y_{jk}^s = y_{kj}^s$

Recall proof of Theorem 2:

$$\rho^\top G \rho = \sum_{(j,k) \in E} (\rho_j - \rho_k)^2 g_{jk}^s + \sum_{j \in \bar{N}} \rho_j^2 \sum_{i: j \sim i} g_{ji}^m > 0$$

Similar structure for strict principal submatrix Y_{22} :

$$\operatorname{Re}(\alpha^H Y_{22} \alpha) = \sum_i \left(\sum_{j,k \in C_i: (j,k) \in E} g_{jk}^s |\alpha_j - \alpha_k|^2 + \sum_{j \in C_i} G_j |\alpha_j|^2 \right)$$

$$\operatorname{Im}(\alpha^H Y_{22} \alpha) = \sum_i \left(\sum_{j,k \in C_i: (j,k) \in E} b_{jk}^s |\alpha_j - \alpha_k|^2 + \sum_{j \in C_i} B_j |\alpha_j|^2 \right)$$

Invertibility of Y_{22}

Derivation

For strict principal submatrix:

$$Y_{22}[j, j] = \sum_{k \notin A: (j, k) \in E} y_{jk}^s + \sum_{k \in A: (j, k) \in E} y_{jk}^s + y_{jj}^m$$

Hence

$$\begin{aligned} \alpha^H Y_{22} \alpha &= \sum_{j \in A} \left(\left(\sum_{k \notin A: (j, k) \in E} y_{jk}^s + \sum_{k \in A: (j, k) \in E} y_{jk}^s + y_{jj}^m \right) |\alpha_j|^2 - \sum_{k \in A: (j, k) \in E} y_{jk}^s \alpha_j^H \alpha_k \right) \\ &= \sum_{j, k \in A: (j, k) \in E} \left(y_{jk}^s |\alpha_j|^2 - y_{jk}^s \alpha_j^H \alpha_k - y_{kj}^s \alpha_k^H \alpha_j + y_{kj}^s |\alpha_k|^2 \right) + \sum_{j \in A} \left(\sum_{k \notin A: (j, k) \in E} y_{jk}^s + y_{jj}^m \right) |\alpha_j|^2 \\ &= \sum_{j, k \in A: (j, k) \in E} y_{jk}^s |\alpha_j - \alpha_k|^2 + \sum_{j \in A} \left(\sum_{k \notin A: (j, k) \in E} y_{jk}^s + y_{jj}^m \right) |\alpha_j|^2 \end{aligned}$$

Invertibility of Y_{22}

Derivation

For strict principal submatrix:

$$Y_{22}[j, j] = \sum_{k \notin A: (j, k) \in E} y_{jk}^s + \sum_{k \in A: (j, k) \in E} y_{jk}^s + y_{jj}^m$$

Hence

$$\operatorname{Re}(\alpha^H Y_{22} \alpha) = \sum_i \left(\sum_{j, k \in C_i: (j, k) \in E} g_{jk}^s |\alpha_j - \alpha_k|^2 + \sum_{j \in C_i} G_j |\alpha_j|^2 \right)$$

$$\operatorname{Im}(\alpha^H Y_{22} \alpha) = \sum_i \left(\sum_{j, k \in C_i: (j, k) \in E} b_{jk}^s |\alpha_j - \alpha_k|^2 + \sum_{j \in C_i} B_j |\alpha_j|^2 \right)$$

Similar conditions to Theorem 2:

$$\rho^T G \rho = \sum_{(j, k) \in E} (\rho_j - \rho_k)^2 g_{jk}^s + \sum_{j \in \bar{N}} \rho_j^2 \sum_{i: j \sim i} g_{ji}^m > 0$$

Invertibility of Y_{22}

When $y_{jk}^s = y_{kj}^s$

Let $y_{jk}^s =: g_{jk}^s + ib_{jk}^s$, $y_{jk}^m =: g_{jk}^m + ib_{jk}^m$, $y_{kj}^m =: g_{kj}^m + ib_{kj}^m$

Conditions

1. For all lines $(j, k) \in E$, $g_{jk}^s \geq 0$; for all buses $j \in \bar{N}$, $G_j \geq 0$
2. For all buses $j \in \bar{N}$, $G_j \neq 0$
3. For all lines $(j, k) \in E$, $g_{jk}^s \neq 0$; for each connected component C_i , $\exists j_i \in C_i$ s.t. $G_{j_i} \neq 0$

Theorem 5

Suppose G is connected and Y is complex symmetric ($y_{jk}^s = y_{kj}^s$). If conditions 1 and either 2 or 3 are satisfied, then

1. $\text{Re}(Y_{22}) \succ 0$
2. Y_{22}^{-1} exists, is symmetric, and $\text{Re}(Y_{22}^{-1}) \succ 0$

Invertibility of Y_{22}

When $y_{jk}^s = y_{kj}^s$

Let $y_{jk}^s =: g_{jk}^s + ib_{jk}^s$, $y_{jk}^m =: g_{jk}^m + ib_{jk}^m$, $y_{kj}^m =: g_{kj}^m + ib_{kj}^m$

Conditions

1. For all lines $(j, k) \in E$, $b_{jk}^s \leq 0$; for all buses $j \in \bar{N}$, $B_j \leq 0$
2. For all buses $j \in \bar{N}$, $B_j \neq 0$
3. For all lines $(j, k) \in E$, $b_{jk}^s \neq 0$; for each connected component C_i , $\exists j_i \in C_i$ s.t. $B_{j_i} \neq 0$

Theorem 6

Suppose G is connected and Y is complex symmetric ($y_{jk}^s = y_{kj}^s$). If conditions 1 and either 2 or 3 are satisfied, then

1. $\text{Im}(Y_{22}) < 0$
2. Y_{22}^{-1} exists, is symmetric, and $\text{Im}(Y_{22}^{-1}) > 0$

Invertibility of Y_{22}

When $y_{jk}^s = y_{kj}^s$ and $y_{jk}^m = y_{kj}^m = 0$

Corollary 7

Suppose G is connected, Y is complex symmetric ($y_{jk}^s = y_{kj}^s$) and $y_{jk}^m = y_{kj}^m = 0$.

1. If $g_{jk}^s > 0$ for all $(j, k) \in E$, then Y_{22}^{-1} exists, is symmetric. Moreover $\text{Re}(Y_{22}) > 0$ and $\text{Re}(Y_{22}^{-1}) > 0$
2. If $b_{jk}^s < 0$ for all $(j, k) \in E$, then Y_{22}^{-1} exists, is symmetric. Moreover $\text{Im}(Y_{22}) < 0$ and $\text{Im}(Y_{22}^{-1}) > 0$

Theorem 8

Suppose G is connected, Y is complex symmetric ($y_{jk}^s = y_{kj}^s$) and $y_{jk}^m = y_{kj}^m = 0$. If $g_{jk}^s \geq 0$ and $b_{jk}^s \leq 0$ $\forall (j, k) \in E$ then

1. $\text{Re}(Y_{22}) \geq 0$, $\text{Im}(Y_{22}) \leq 0$, $\text{Re}(Y_{22}) - \text{Im}(Y_{22}) > 0$
2. Y_{22}^{-1} exists and is symmetric

Invertibility of Y/Y_{22}

When $y_{jk}^s = y_{kj}^s$

Theorem 9

Suppose Y_{22} is nonsingular.

1. If $\operatorname{Re}(Y) > 0$, then $(Y/Y_{22})^{-1}$ exists and is symmetric. Moreover $\operatorname{Re}(Y/Y_{22}) > 0$ and $\operatorname{Re}\left((Y/Y_{22})^{-1}\right) > 0$
2. If $\operatorname{Im}(Y) < 0$, then $(Y/Y_{22})^{-1}$ exists and is symmetric. Moreover $\operatorname{Im}(Y/Y_{22}) < 0$ and $\operatorname{Im}\left((Y/Y_{22})^{-1}\right) > 0$

Outline

1. Component models
2. Network model: VI relation
 - Example and network model
 - Admittance matrix Y and properties
 - Kron reduction Y/Y_{22} and properties
 - Radial network
3. Network model: Vs relation
4. Computation methods
5. Linear power flow model

Radial networks

When $y_{jk}^s = y_{kj}^s$ and $y_{jk}^m = y_{kj}^m = 0$

$(N + 1) \times N$ incidence matrix C , $D_y^s := \text{diag}(y_l^s, l \in E)$:

$$Y = CD_y^s C^T \quad \text{admittance matrix}$$

$N \times N$ **reduced** incidence matrix \hat{C} , $D_y^s := \text{diag}(y_l^s, l \in E)$:

$$\hat{Y} = \hat{C}D_y^s \hat{C}^T \quad \text{reduced admittance matrix}$$

Main property: \hat{C} and hence \hat{Y} are always nonsingular. Moreover $\hat{Z} := \hat{Y}^{-1}$ has a simple and useful structure

Radial networks

When $y_{jk}^s = y_{kj}^s$ and $y_{jk}^m = y_{kj}^m = 0$

Theorem 10

Suppose G is connected, Y is complex symmetric ($y_{jk}^s = y_{kj}^s$) and $y_{jk}^m = y_{kj}^m = 0$.

1. Reduced incidence matrix \hat{C} is nonsingular

$$[\hat{C}^{-1}]_{lj} = \begin{cases} -1 & l \in P_j \\ 1 & -l \in P_j \\ 0 & \text{otherwise} \end{cases}$$

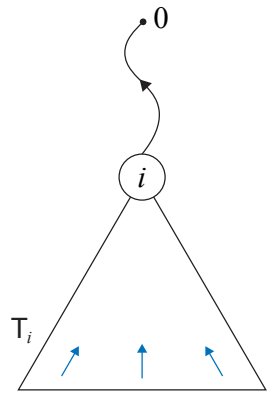
2. Reduced admittance matrix \hat{Y} is nonsingular, and

$$\hat{Z} := \hat{Y}^{-1} = \hat{C}^{-T} D_z^s \hat{C}^{-1}$$

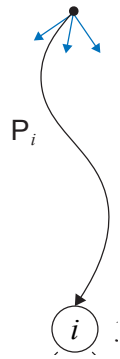
$$\hat{Z}_{jk} = \sum_{l \in P_j \cap P_k} z_l^s$$

sum of $z_{jk}^s := 1/y_{jk}^s$ on common segment of paths from ref bus 0 to j and k

This property has been applied for topology identification, voltage control, ...



T_i : subtree rooted at bus i



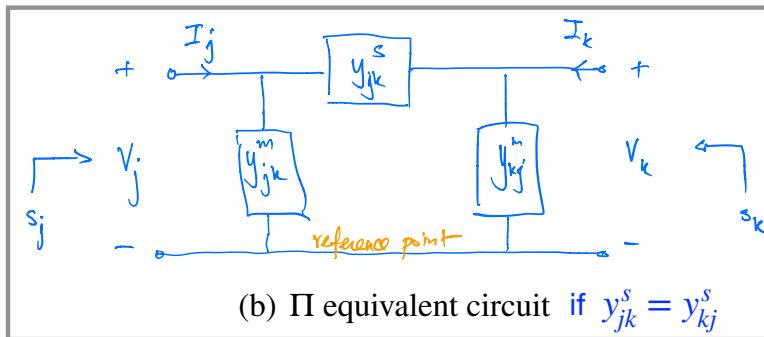
T_i : unique path from 0 to i

Outline

1. Component models
2. Network model: IV relation
3. Network model: sV relation
 - Complex form
 - Polar form
 - Cartesian form
 - Types of buses
 - Application: topology identification
4. Computation methods
5. Linear power flow model

General network

Branch currents



Sending-end currents

$$I_{jk} = y_{jk}^s(V_j - V_k) + y_{jk}^m V_j$$

$$I_{kj} = y_{kj}^s(V_k - V_j) + y_{kj}^m V_k$$

Power flow models

Complex form

Using $S_{jk} := V_j I_{jk}^H$:

$$S_{jk} = \left(y_{jk}^s\right)^H \left(|V_j|^2 - V_j V_k^H\right) + \left(y_{jk}^m\right)^H |V_j|^2$$

$$S_{kj} = \left(y_{kj}^s\right)^H \left(|V_k|^2 - V_k V_j^H\right) + \left(y_{kj}^m\right)^H |V_k|^2$$

Power flow models

Complex form

Bus injection model $s_j = \sum_{k:j \sim k} S_{jk}$:

$$s_j = \sum_{k:j \sim k} \left(y_{jk}^s \right)^H \left(|V_j|^2 - V_j V_k^H \right) + \left(y_{jj}^m \right)^H |V_j|^2$$

In terms of admittance matrix Y

$$s_j = \sum_{k=1}^{N+1} Y_{jk}^H V_j V_k^H$$

$N + 1$ complex equations in $2(N + 1)$ complex variables $\left(s_j, V_j, j \in \bar{N} \right)$

Power flow models

Polar form

Write $s_j =: p_j + iq_j$ and $V_j =: |V_j| e^{i\theta_j}$ with $y_{jk}^s =: g_{jk}^s + ib_{jk}^s$, $y_{jk}^m =: g_{jk}^m + ib_{jk}^m$:

$$p_j = \sum_{k:k\sim j} (g_{jk}^s + g_{jk}^m) |V_j|^2 - \sum_{k:k\sim j} |V_j| |V_k| (g_{jk}^s \cos \theta_{jk} + b_{jk}^s \sin \theta_{jk})$$

$$q_j = - \sum_{k:k\sim j} (b_{jk}^s + b_{jk}^m) |V_j|^2 - \sum_{k:k\sim j} |V_j| |V_k| (g_{jk}^s \sin \theta_{jk} - b_{jk}^s \cos \theta_{jk})$$

$2(N + 1)$ real equations in $4(N + 1)$ real variables $(p_j, q_j, |V_j|, \theta_j, j \in \bar{N})$

Power flow models

Cartesian form

Write $s_j =: p_j + iq_j$ and $V_j =: c_j + id_j$ with $c_j = |V_j| \cos \theta_j$ and $d_j = |V_j| \sin \theta_j$:

$$p_j = \sum_{k:k \sim j} \left(g_{jk}^s + g_{jk}^m \right) \left(c_j^2 + d_j^2 \right) - \sum_{k:k \sim j} \left(g_{jk}^s (c_j c_k + d_j d_k) + b_{jk}^s (d_j c_k - c_j d_k) \right)$$

$$q_j = - \sum_{k:k \sim j} \left(b_{jk}^s + b_{jk}^m \right) \left(c_j^2 + d_j^2 \right) - \sum_{k:k \sim j} \left(g_{jk}^s (d_j c_k - c_j d_k) - b_{jk}^s (c_j c_k + d_j d_k) \right)$$

$2(N + 1)$ real equations in $4(N + 1)$ real variables $(p_j, q_j, c_j, d_j, j \in \bar{N})$

Power flow models

Types of buses

Power flow equations specify $2(N + 1)$ real equations in $4(N + 1)$ real variables

- Power flow (load flow) problem: given $2(N + 1)$ values, determine remaining vars

Types of buses

- *PV* buses : $(p_j, |V_j|)$ specified, determine (q_j, θ_j) , e.g. generator
- *PQ* buses : (p_j, q_j) specified, determine V_j , e.g. load
- Slack bus 0 : $V_0 := 1 \angle 0^\circ$ pu specified, determine (p_0, q_0)

Outline

1. Component models
2. Network model: IV relation
3. Network model: sV relation
4. Computation methods
 - Gauss-Seidel algorithm
 - Newton-Raphson algorithm
 - Fast decoupled algorithm
5. Linear power flow model

Computation methods

Gauss-Seidel algorithm

Case 1: given V_0 and (s_1, \dots, s_N) , determine s_0 and (V_1, \dots, V_N)

Power flow equations

$$s_0 = \sum_k Y_{0k}^H V_0 V_k^H$$
$$s_j = \sum_k Y_{jk}^H V_j V_k^H, \quad j \in N$$

- First compute (V_1, \dots, V_N)
- Then compute s_0

Computation methods

Gauss-Seidel algorithm

Case 1: given V_0 and (s_1, \dots, s_N) , determine s_0 and (V_1, \dots, V_N)

Rearrange 2nd equation:

$$\frac{s_j^H}{V_j^H} = Y_{jj}V_j + \sum_{\substack{k=0 \\ k \neq j}}^N Y_{jk}V_k, \quad j \in N$$

$$V_j = \frac{1}{Y_{jj}} \left(\frac{s_j^H}{V_j^H} - \sum_{\substack{k=0 \\ k \neq j}}^N Y_{jk}V_k \right) =: f_j(V_1, \dots, V_N), \quad j \in N$$

Computation methods

Gauss-Seidel algorithm

Case 1: given V_0 and (s_1, \dots, s_N) , determine s_0 and (V_1, \dots, V_N)

2nd power flow equation:

$$V = f(V)$$

where $V := (V_j, j \in N)$, $f := (f_j, j \in N)$

Gauss algorithm is the fixed point iteration

$$V(t+1) = f(V(t))$$

Computation methods

Gauss-Seidel algorithm

Case 1: given V_0 and (s_1, \dots, s_N) , determine s_0 and (V_1, \dots, V_N)

Gauss algorithm:

$$V_1(t+1) = f_1(V_1(t), \dots, V_N(t))$$

$$V_2(t+1) = f_2(V_1(t), \dots, V_N(t))$$

\vdots

$$V_N(t+1) = f_N(V_1(t), \dots, V_{N-1}(t), V_N(t))$$

Computation methods

Gauss-Seidel algorithm

Case 1: given V_0 and (s_1, \dots, s_N) , determine s_0 and (V_1, \dots, V_N)

Gauss-Seidel algorithm:

$$V_1(t+1) = f_1(V_1(t), \dots, V_N(t))$$

$$V_2(t+1) = f_2(V_1(t+1), \dots, V_N(t))$$

\vdots

$$V_N(t+1) = f_N(V_1(t+1), \dots, V_{N-1}(t+1), V_N(t))$$

Computation methods

Gauss-Seidel algorithm

Case 2: given (V_0, \dots, V_m) and (s_{m+1}, \dots, s_N) , determine $(s_j, j \leq m)$ and $(V_j, j > m)$

Power flow equations

$$s_j = \sum_k Y_{jk}^H V_j V_k^H, \quad j \leq m$$

$$s_j = \sum_k Y_{jk}^H V_j V_k^H, \quad j > m$$

- First compute (V_{m+1}, \dots, V_N) from 2nd set of equations using the same algorithm
- Then compute $(s_j, j \leq m)$ from 1st set of equations

Computation methods

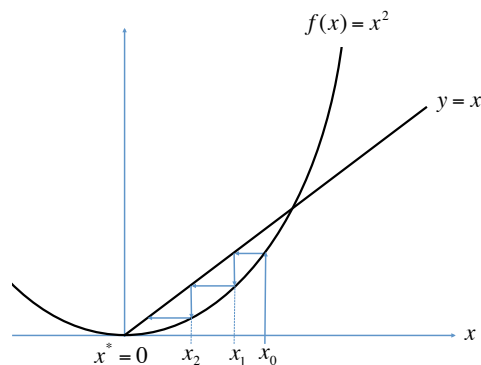
Gauss-Seidel algorithm

If algorithm converges, the limit is a fixed point and a power flow solution

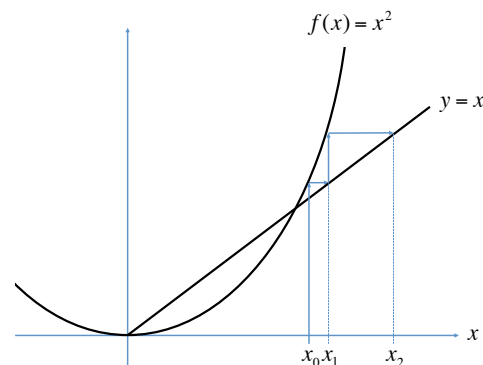
Algorithm converges linearly to unique fixed point if f is a contraction mapping

- Contraction is sufficient, but not necessary, for convergence

In general, algorithm may or may not converge depending on initial point



(a) Convergence



(b) Divergence

Computational methods

Newton-Raphson algorithm

To solve $f(x) = 0$

where $f: \mathbb{R}^n \rightarrow \mathbb{R}^n$, e.g. $\nabla F(x) = 0$ for unconstrained optimization

Idea:

- Linear approximation

$$\hat{f}(x(t+1)) = f(x(t)) + J(x(t)) \Delta x(t)$$

- Choose $\Delta x(t)$ such that $\hat{f}(x(t+1)) = 0$, i.e., solve

$$J(x(t)) \Delta x(t) = -f(x(t))$$

- Next iterate $x(t+1) := x(t) + \Delta x(t)$

$$J(x) := \frac{\partial f}{\partial x}(x) = \begin{bmatrix} \frac{\partial f_1}{\partial x_1}(x) & \cdots & \frac{\partial f_1}{\partial x_n}(x) \\ \vdots & \vdots & \vdots \\ \frac{\partial f_n}{\partial x_1}(x) & \cdots & \frac{\partial f_n}{\partial x_n}(x) \end{bmatrix}$$

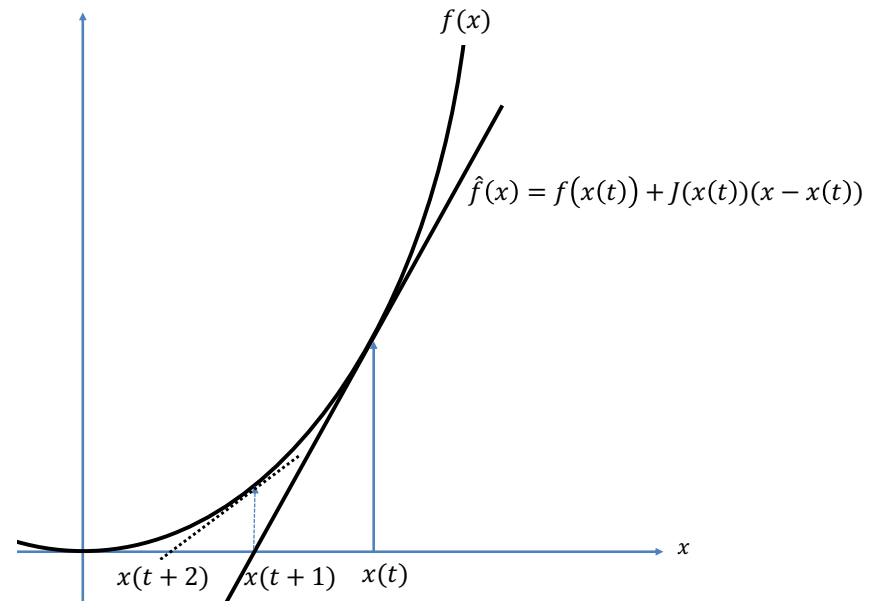
Computational methods

Newton-Raphson algorithm

To solve $f(x) = 0$

where $f: \mathbb{R}^n \rightarrow \mathbb{R}^n$, e.g. $\nabla F(x) = 0$ for unconstrained optimization

$$x(t+1) := x(t) - (J(x(t)))^{-1} f(x(t))$$



Computational methods

Newton-Raphson algorithm

Kantorovic Theorem

Consider $f: D \rightarrow \mathbb{R}^n$ where $D \subseteq \mathbb{R}^n$ is an open convex set. Suppose

- f is differentiable and ∇f is Lipschitz on D , i.e., $\|\nabla f(y) - \nabla f(x)\| \leq L\|y - x\|$
- $x_0 \in D$ and $\nabla f(x_0)$ is invertible

Let $\beta \geq \left\| (\nabla f(x_0))^{-1} \right\|$, $\eta \geq \left\| (\nabla f(x_0))^{-1} f(x_0) \right\|$ and

$$h := \beta\eta L, \quad r := \frac{1 - \sqrt{1 - 2h}}{h} \eta$$

Computational methods

Newton-Raphson algorithm

Kantorovic Theorem

Consider $f: D \rightarrow \mathbb{R}^n$ where $D \subseteq \mathbb{R}^n$ is an open convex set. Suppose

- f is differentiable and ∇f is Lipschitz on D , i.e., $\|\nabla f(y) - \nabla f(x)\| \leq L\|y - x\|$
- $x_0 \in D$ and $\nabla f(x_0)$ is invertible

If the closed ball $B_r(x_0) \subseteq D$ and $h \leq 1/2$, then Newton iteration

$$x(t+1) := x(t) - (\nabla f(x(t)))^{-1} f(x(t))$$

converges to a solution $x^* \in B_r(x_0)$ of $f(x) = 0$

Newton-Raphson converges if it starts close to a solution, often quadratically

Computational methods

Newton-Raphson algorithm

Apply to power flow equations in polar form:

$$p_j(\theta, |V|) = p_j, \quad j \in N$$

$$q_j(\theta, |V|) = q_j, \quad j \in N_{pq}$$

where

$$p_j(\theta, |V|) := \left(\sum_{k=0}^N g_{jk} \right) |V_j|^2 - \sum_{k \neq j} |V_j| |V_k| \left(g_{jk} \cos \theta_{jk} + b_{jk} \sin \theta_{jk} \right)$$

$$q_j(\theta, |V|) := - \left(\sum_{k=0}^N b_{jk} \right) |V_j|^2 - \sum_{k \neq j} |V_j| |V_k| \left(g_{jk} \sin \theta_{jk} - b_{jk} \cos \theta_{jk} \right)$$

Computational methods

Newton-Raphson algorithm

Define $f : \mathbb{R}^{N+N_{qp}} \rightarrow \mathbb{R}^{N+N_{qp}}$

$$f(\theta, |V|) := \begin{bmatrix} \Delta p(\theta, |V|) \\ \Delta q(\theta, |V|) \end{bmatrix} := \begin{bmatrix} p(\theta, |V|) - p \\ q(\theta, |V|) - q \end{bmatrix}$$

with

$$J(\theta, |V|) := \begin{bmatrix} \frac{\partial p}{\partial \theta} & \frac{\partial p}{\partial |V|} \\ \frac{\partial q}{\partial \theta} & \frac{\partial q}{\partial |V|} \end{bmatrix}$$

Computational methods

Newton-Raphson algorithm

1. Initialization: choose $(\theta(0), |V(0)|)$
2. Iterate until stopping criteria

(a) Determine $(\Delta\theta(t), \Delta|V|(t))$ from

$$J(\theta(t), |V|(t)) \begin{bmatrix} \Delta\theta(t) \\ \Delta|V|(t) \end{bmatrix} = - \begin{bmatrix} \Delta p(\theta(t), |V|(t)) \\ \Delta q(\theta(t), |V|(t)) \end{bmatrix}$$

(b) Set

$$\begin{bmatrix} \theta(t+1) \\ |V|(t+1) \end{bmatrix} := \begin{bmatrix} \theta(t) \\ |V|(t) \end{bmatrix} + \begin{bmatrix} \Delta\theta(t) \\ \Delta|V|(t) \end{bmatrix}$$

Computational methods

Fast Decoupled algorithm

Key observation: the Jacobian is roughly block-diagonal

$$J(\theta, |V|) := \begin{bmatrix} \frac{\partial p}{\partial \theta} & \frac{\partial p}{\partial |V|} \\ \frac{\partial q}{\partial \theta} & \frac{\partial q}{\partial |V|} \end{bmatrix} \approx \begin{bmatrix} \frac{\partial p}{\partial \theta} & 0 \\ 0 & \frac{\partial q}{\partial |V|} \end{bmatrix}$$

i.e., decoupling between p and $|V|$, and between q and θ

Computational methods

Fast Decoupled algorithm

Key observation: the Jacobian is roughly block-diagonal

$$J(\theta, |V|) := \begin{bmatrix} \frac{\partial p}{\partial \theta} & \frac{\partial p}{\partial |V|} \\ \frac{\partial q}{\partial \theta} & \frac{\partial q}{\partial |V|} \end{bmatrix} \approx \begin{bmatrix} \frac{\partial p}{\partial \theta} & 0 \\ 0 & \frac{\partial q}{\partial |V|} \end{bmatrix}$$

i.e., decoupling between p and $|V|$, and between q and θ

This simplifies the computation of $(\Delta\theta(t), \Delta|V|(t))$

$$\frac{\partial p}{\partial \theta}(\theta(t), |V|(t)) \Delta\theta(t) = -\Delta p(\theta(t), |V|(t))$$

$$\frac{\partial q}{\partial |V|}(\theta(t), |V|(t)) \Delta|V|(t) = -\Delta q(\theta(t), |V|(t))$$

Computational methods

Fast Decoupled algorithm

Decoupling assumption: $g_{jk} = 0, \sin \theta_{jk} = 0$

$$\frac{\partial p_j}{\partial |V_k|} = \begin{cases} -|V_j| \left(g_{jk} \cos \theta_{jk} + b_{jk} \sin \theta_{jk} \right), & j \neq k \\ \frac{p_j(\theta, |V|)}{|V_j|} + \left(\sum_i g_{ji} \right) |V_j|, & j = k \end{cases}$$

$$g_{jk} = 0, \sin \theta_{jk} = 0, p_j(\theta, |V|) = 0 \Rightarrow \frac{\partial p}{\partial |V|} = 0$$

Computational methods

Fast Decoupled algorithm

Decoupling assumption: $g_{jk} = 0, \sin \theta_{jk} = 0$

$$\frac{\partial q_j}{\partial \theta_k} = \begin{cases} |V_j| |V_k| (g_{jk} \cos \theta_{jk} + b_{jk} \sin \theta_{jk}), & j \neq k \\ p_j(\theta, |V|) - \left(\sum_i g_{ji} \right) |V_j|^2, & j = k \end{cases}$$

$$g_{jk} = 0, \sin \theta_{jk} = 0, p_j(\theta, |V|) = 0 \Rightarrow \frac{\partial q}{\partial \theta} = 0$$

Outline

1. Component models
2. Network model: IV relation
3. Network model: sV relation
4. Computation methods
5. Linear power flow model
 - Laplacian matrix L
 - DC power flow model

Laplacian matrix L

Given a graph $G := (V, E)$ with $n \times m$ node-by-line incidence matrix C and line susceptances $B := \text{diag}(b_l, l \in E)$, the **Laplacian matrix** is

$$L := CBC^T$$

Assumptions:

- L is real symmetric
- All row and column sums are zero
- $b_l > 0$ for all $l \in E$

Lemma

For all $x \in \mathbb{R}^n$, $x^T L x = \sum_{(j,k) \in E} b_{jk} (x_j - x_k)^2 \geq 0$

Proof: $x^T L x = \sum_j \sum_k L_{jk} x_j x_k = \sum_{(i,j) \in E} b_{ij} (x_i^2 - 2x_i x_j + x_j^2) = \sum_{(i,j) \in E} b_{ij} (x_i - x_j)^2$

Laplacian matrix L

Theorem

Suppose G contains $K \geq 1$ connected components.

1. L is positive semidefinite
2. $\text{rank}(L) = N - K$ with $\text{null}(L) = \{x : x_j = x_k, \forall j, k \in \text{each connected component}\}$
3. Suppose $K = 1$. Then

- $\text{rank}(L) = n - 1$ with $\text{null}(L) = \text{span}(\mathbf{1})$

- Pseudo-inverse of L is $L^\dagger = \left(L + \frac{1}{n}\mathbf{1}\mathbf{1}^\top\right)^{-1} - \frac{1}{n}\mathbf{1}\mathbf{1}^\top = \sum_{j=2}^n \frac{1}{\lambda_j} v_j v_j^\top$

- Both L and L^\dagger are symmetric and have zero row (and column) sums

- $LL^\dagger = L^\dagger L = \mathbb{I}_n - \frac{1}{n}\mathbf{1}\mathbf{1}^\top$

- For x with $\mathbf{1}^\top x = 0$, $LL^\dagger x = L^\dagger Lx = x$

Laplacian matrix L

Theorem

4. Suppose $K = 1$. Then

- Any $k \times k$ principal submatrix M of L is nonsingular for $k \leq n - 1$
- Both M and M^{-1} are symmetric

in contrast to **complex** symmetric admittance matrix $Y = CD_y^s C^T$ whose submatrix Y_{22} may be singular

Laplacian matrix L

Summary: comparison

Invertibility of admittance matrices:

1. Complex symmetric Y
 - A strict principal submatrix Y_{22} is not always nonsingular
 - Y_{22} is nonsingular if $\text{Re}(Y) \succ 0$ or if $\text{Im}(Y) \prec 0$
2. Complex symmetric Y for connected radial network
 - \hat{Y} corresponding to removing any leaf node is always nonsingular
 - Any strict principal submatrix Y_{22} corresponding to a (connected) subtree is always nonsingular (by induction)
3. Real symmetric Laplacian matrix L with zero row sums and $B \succ 0$
 - Any strict principal submatrix M is nonsingular

DC power flow model

Consider power network modeled by a connected graph $G := (\bar{N}, E)$ with $N + 1$ buses and M lines

Assumptions

- Lossless: series conductances $\tilde{g}_l^s = 0$, shunt admittances $\tilde{y}_{jk}^m = \tilde{y}_{kj}^m = 0$; $\tilde{b}_{jk}^s < 0$
- Small angle differences: $\sin(\theta_j - \theta_k) \approx \theta_j - \theta_k$
- Voltage magnitudes $|V_j|$ are fixed and given
- Ignore reactive power

These assumptions are reasonable for transmission networks (not for distribution networks)

Substituting directly into polar form power flow equation yields

$$p_j = \sum_{k:j \sim k} \left(-\tilde{b}_{jk}^s |V_j| |V_k| \right) (\theta_j - \theta_k) \quad =: \quad \sum_{k:j \sim k} b_{jk} (\theta_j - \theta_k), \quad b_{jk} > 0$$

(When $|V_j| = \mu, \forall j$, DC power flow is also linearization of polar form power flow equation around flat voltage profile)

DC power flow model

In vector form

Let

- $C : (N + 1) \times M$ incidence matrix
- $B := \text{diag}(b_l, l \in E) \succ 0$
- P : line flow (M -vector)

DC power flow model:

$$p = CP, \quad P = BC^T\theta$$

Eliminate $P \implies p = CBC^T\theta =: L\theta$

Given p with $\mathbf{1}^T p = 0$ (power balance), solution: $P = BC^T L^\dagger p, \quad \theta = L^\dagger p + a\mathbf{1}$

These are equivalent specification of DC power flow model

DC power flow model

In vector form

Remarks

- $\mathbf{1}^\top p = \mathbf{1}^\top CP = 0$: generation = demand, due to lossless assumption
- $\theta = L^\dagger p + a\mathbf{1}$: arbitrary constant a can be fixed by choosing a reference node, e.g., $\theta_0 := 0$
- P : line flow (M -vector)
- Most of DC power flow properties (as well as DC OPF, PTDF, LODF, ...) originates from properties of Laplacian matrix L

DC power flow model

In terms of \hat{L}^{-1}

Remarks

Let

- \hat{C}, \hat{L} : the **reduced** incidence matrix and **reduced** Laplacian matrix respectively
- $(\hat{p}, \hat{\theta})$: power injections and voltage angles at non-reference buses

Then \hat{L}^{-1} exists

Given arbitrary \hat{p} at non-reference buses, power flow solution is often expressed in terms of \hat{L}^{-1} in the literature:

$$P = BC^T \hat{L}^{-1} \hat{p}, \quad \hat{\theta} = \hat{L}^{-1} \hat{p}$$

c.f. $P = BC^T L^\dagger p, \quad \theta = L^\dagger p + a\mathbf{1}$

This solution is unique and assumes $\theta_0 := 0$ at bus 0.

This model is a special case of the solution in terms of the pseudo-inverse L^\dagger with a s.t. $\theta_0 := 0$, and therefore less flexible because \hat{L} depends on the choice of reference bus

DC power flow model

In terms of \hat{L}^{-1}

Lemma

$$P = BC^T \hat{L}^{-1} \hat{p} = BC^T L^\dagger p, \quad \hat{\theta} = \hat{L}^{-1} \hat{p}$$

i.e. line flows P are independent of choice of reference bus or \hat{L}

This result can be generalized to the case where price reference (slack) bus r ($p_r = -\mathbf{1}^T p_{-r}$) and angle reference bus 0 ($\theta_0 := 0$) are different

- Optimal dispatch and locational marginal prices are independent of the choice of (angle or price) reference bus
- It is easier however to use L^\dagger instead of \hat{L}

Summary

1. Component models
 - Single-phase devices, line, transformer
2. Network models
 - IV relation (admittance matrix Y), sV relation (power flow equations)
 - Radial network: inverse of reduced admittance matrix has simple structure
3. Computation methods
 - Gauss-Seidel algorithm, Newton-Raphson algorithm, Fast decoupled algorithm
4. Linear power flow models
 - Laplacian matrix L , DC power flow model